

Interface State Analysis on Nonsilicon Semiconductors and The Role of Heterostructures

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Nonsilicon Interfaces - Outline

Introduction to Nonsilicon Channel Materials

Quiz: Do You Understand Nonsilicon Interface Data?

Interface State Analysis

- *Introduction*
- *Admittance-voltage (capture/emission):*
 - *Flow chart for trap evaluation*
 - *Example D_{it} quantification*
- *Photoluminescence intensity (recombination)*

D_{it} and Heterostructures in MOSFETs

- *Heterostructure barrier layer: Introduction*
- *Shifting D_{it} distribution in energy space*
- *D_{it} amplification or reduction – Barrier layer case studies*

Quiz Solution

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Quiz Solution

Introduction to Nonsilicon Channels

Material	Lattice constant	Bandgap	Electron injection velocity @ 0.5 V
Silicon	0.5431 nm	1.12 eV	0.5×10^7 cm/s
GaAs	0.5653 nm	1.42 eV	1.5×10^7 cm/s
$\text{In}_{0.5}\text{Ga}_{0.5}\text{As}$	0.5869 nm	0.74 eV	2.7×10^7 cm/s
InAs	0.6058 nm	0.35 eV	3.7×10^7 cm/s
InSb	0.6479 nm	0.17 eV	4.6×10^7 cm/s

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Interface State Analysis: Introduction

Many measurement techniques, among them

- Admittance-voltage: Capacitance-voltage (C-V)
high/low frequency, low frequency, Terman method ...
- Admittance-voltage: Conductance-voltage (G-V)
- Charge Pumping
- Photoluminescence intensity (PL-I) ...

Physics

- Capture/emission of charge carrier at trapping center
- Recombination of charge carriers at trapping center

Major Parameters

- Bandstructure (bandgap, direct, indirect ...)
- $0.35 \text{ eV} \leq \text{bandgap} \leq 1.42 \text{ eV}$

Interface State Analysis: Introduction

Many measurement techniques, among them

- Admittance-voltage: **Capacitance-voltage (C-V)**
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- Admittance-voltage: Conductance-voltage (G-V)
- Charge Pumping
- **Photoluminescence intensity (PL-I) ...**

Physics

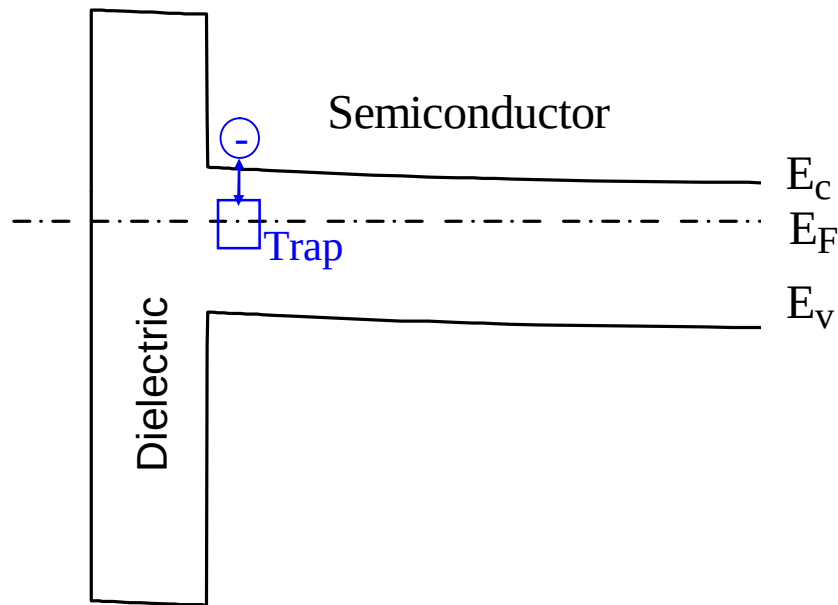
- **Capture/emission of charge carrier at trapping center**
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Major Parameters

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Interface State Analysis: Introduction

Carrier Capture/Emission τ



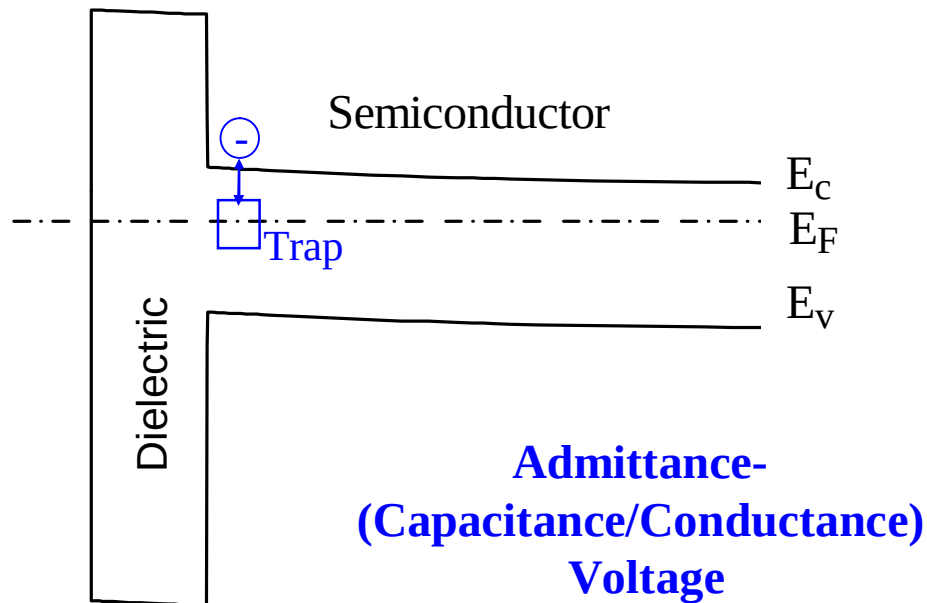
- τ is exponential function of E_T
- Interface state density $D_{it} = f(E_T)$

Interface State Analysis: Introduction



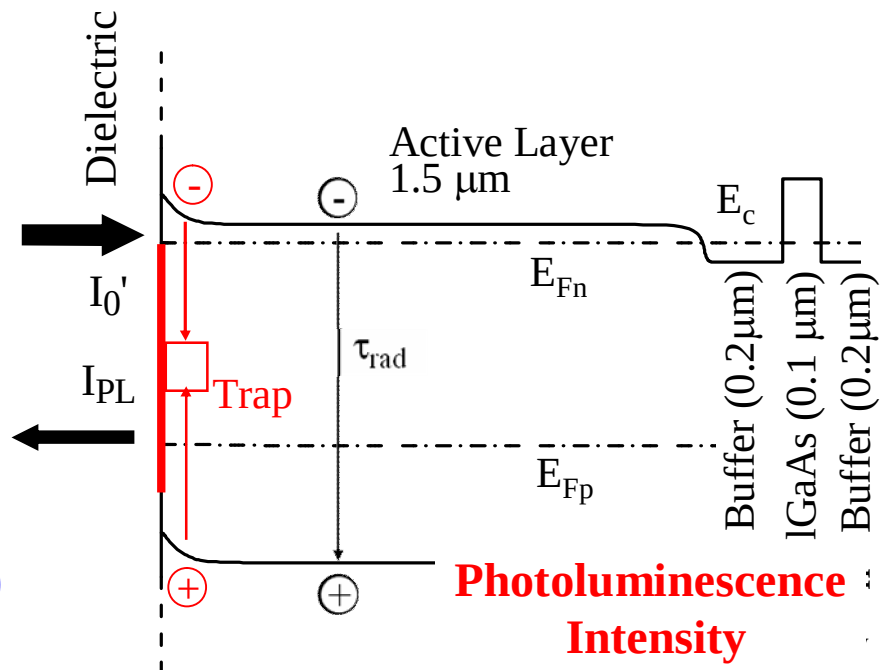
TSMC Property

Carrier Capture/Emission τ



- τ is exponential function of E_T
- Interface state density $D_{it} = f(E_T)$

Carrier Recombination $\tau_{n/p}$



- $\tau_{n/p} \neq f(\text{trap energy } E_T)$
- D_{it} integrated over bandgap

Admittance-Voltage Techniques: Examples

High/low frequency C-V

- Traps must follow DC sweep (quasi-equilibrium)
- Traps must follow low frequency (quasi-static) signal
- Traps must not follow high frequency signal

Low frequency C-V

- Traps must follow DC sweep (quasi-equilibrium)
- Traps must follow low frequency (quasi-static) signal
- Advanced model required to extract D_{it}

Terman Method (C-V)

- Traps must follow DC sweep (quasi-equilibrium)
- Traps must not follow ac signal

G-V Method

- Traps must follow DC signal (quasi-equilibrium)
- Traps must partially follow ac signal

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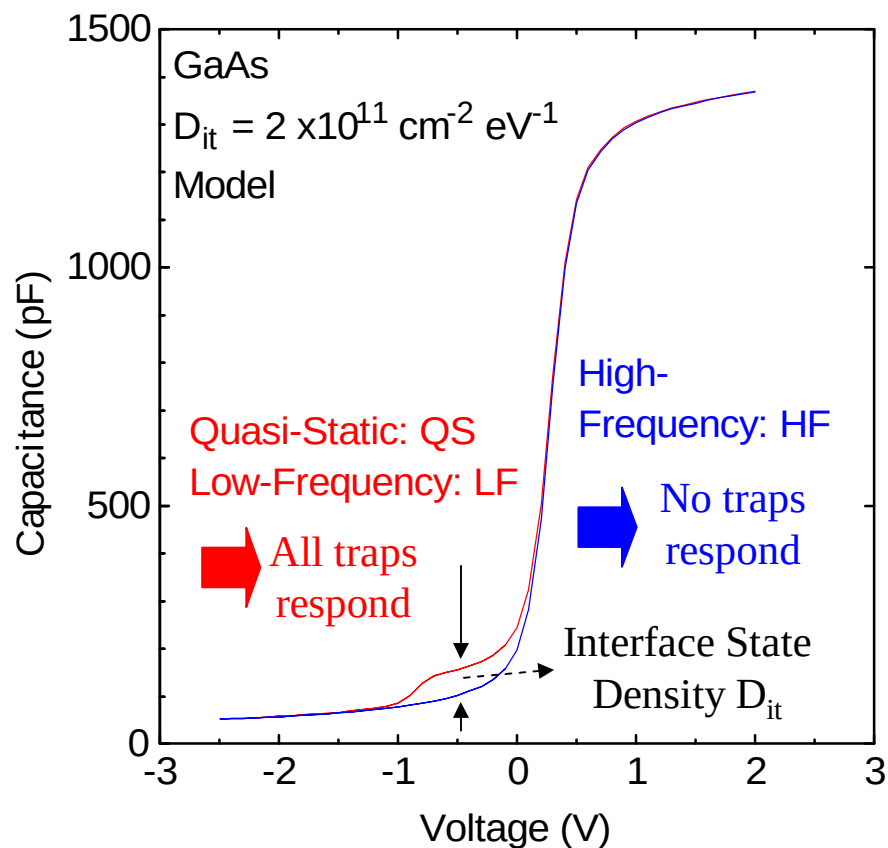
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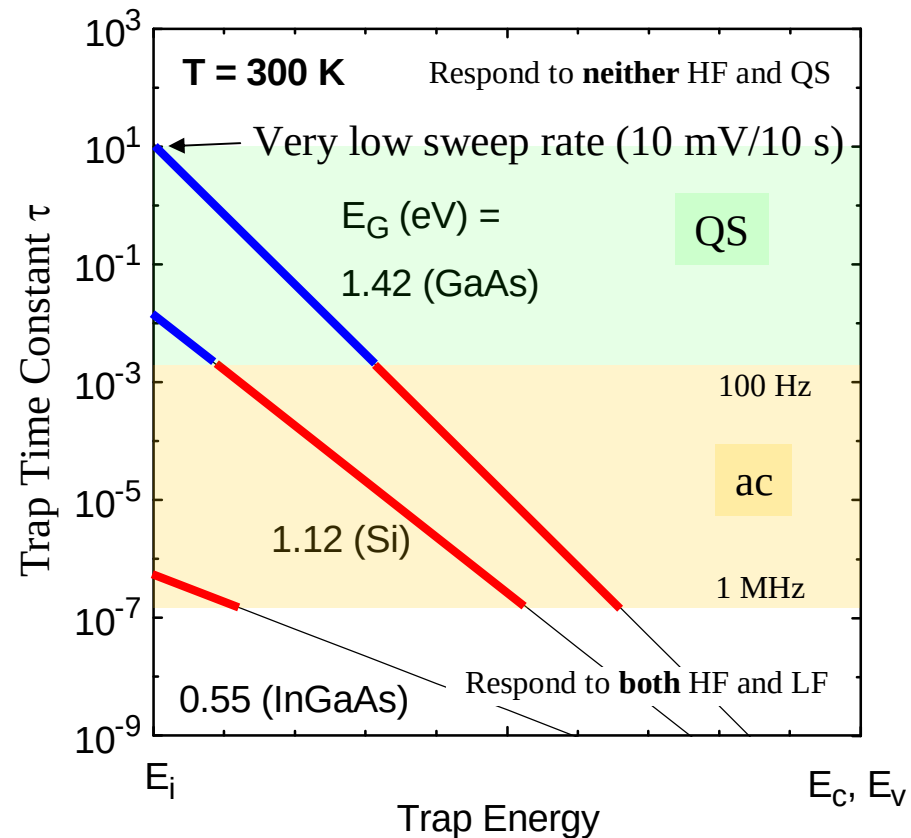
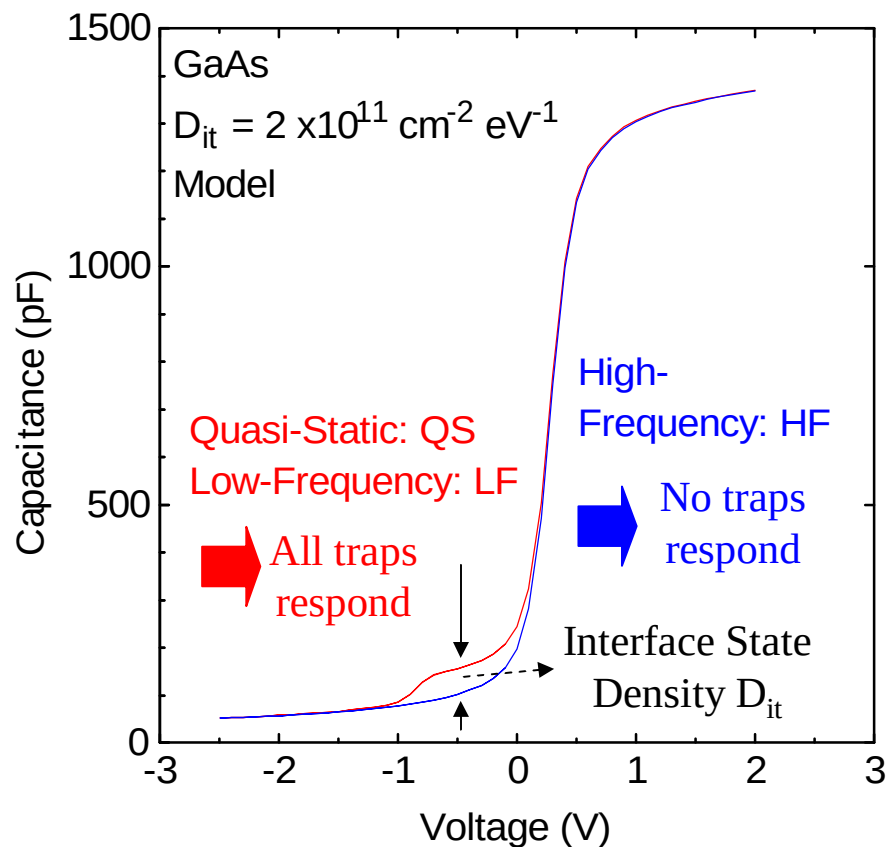
G-V Method

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Example C-V Analysis: High/Low Frequency

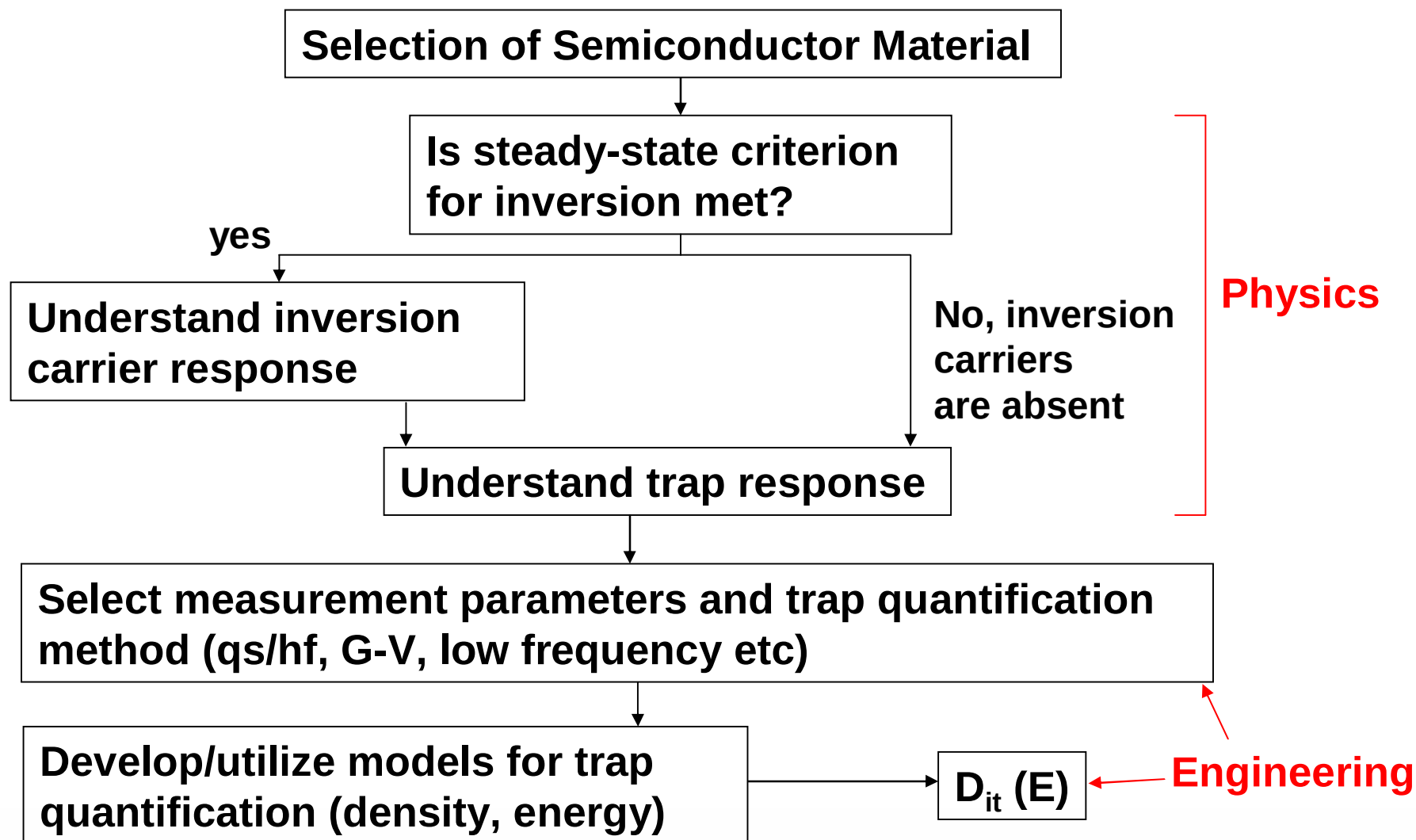


Example C-V Analysis: High/Low Frequency

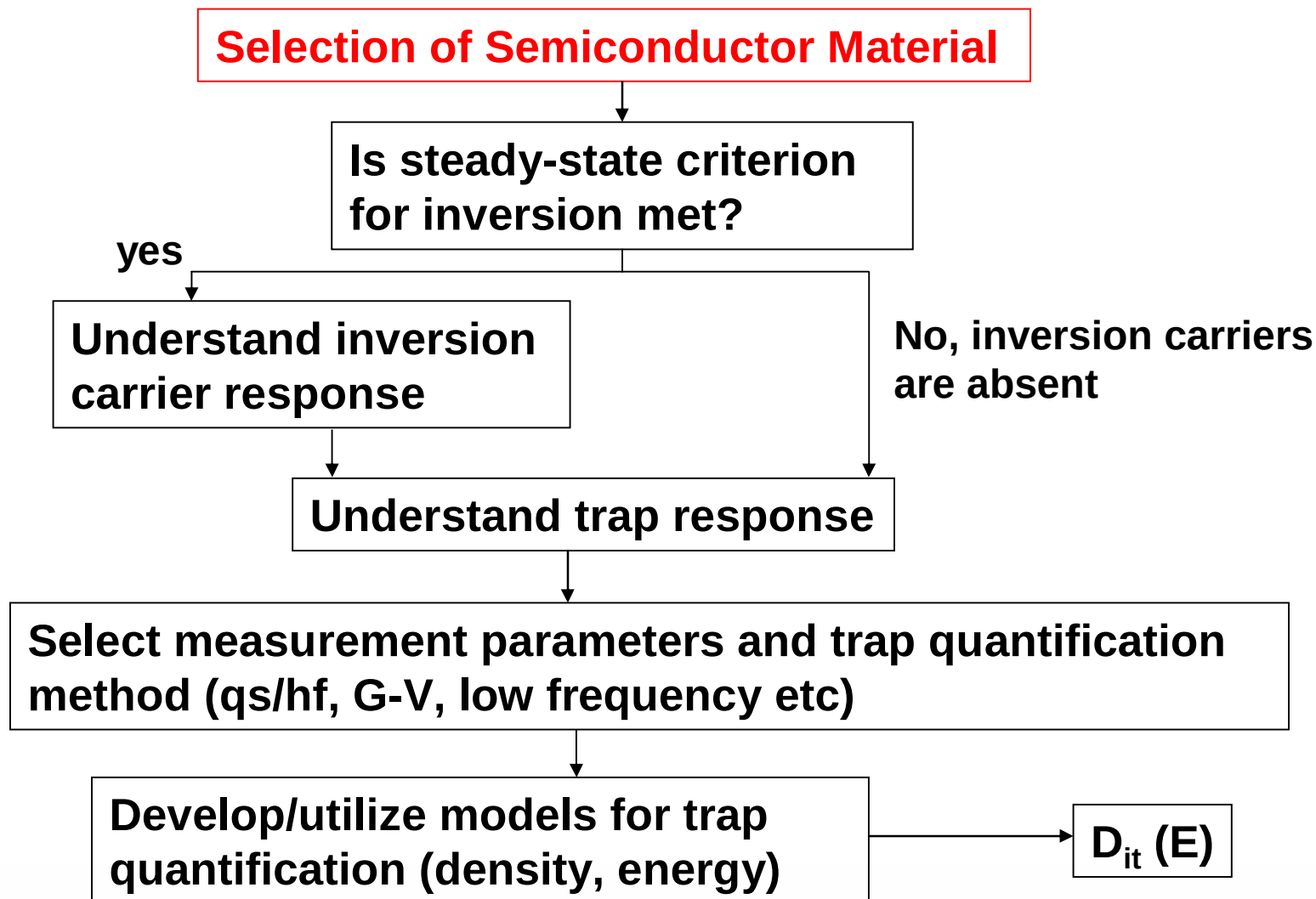


When changing the bandgap, MOS physics needs to be investigated and measurement parameters adjusted accordingly

Admittance-Voltage: Flow Chart



Admittance-Voltage: Flow Chart



Selection of Semiconductor Material

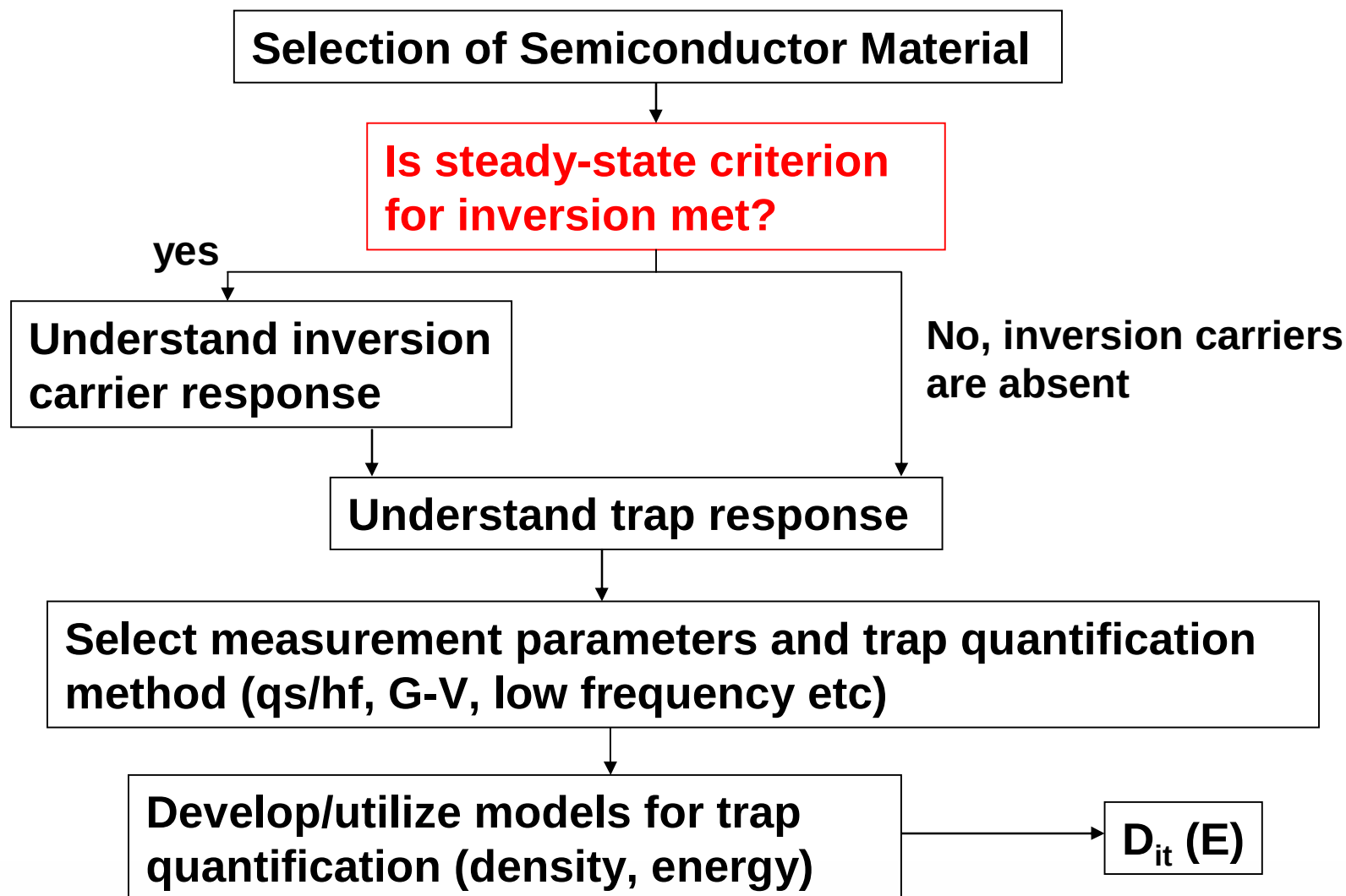
Very important because:

- *Trap time constants vary over many orders of magnitude as a function of bandgap, energy position, and temperature*
- *Measurement techniques need to be properly selected*
- *Measurement parameters need to be adjusted*
- *Models for trap quantification need to be suitable*

In this tutorial:

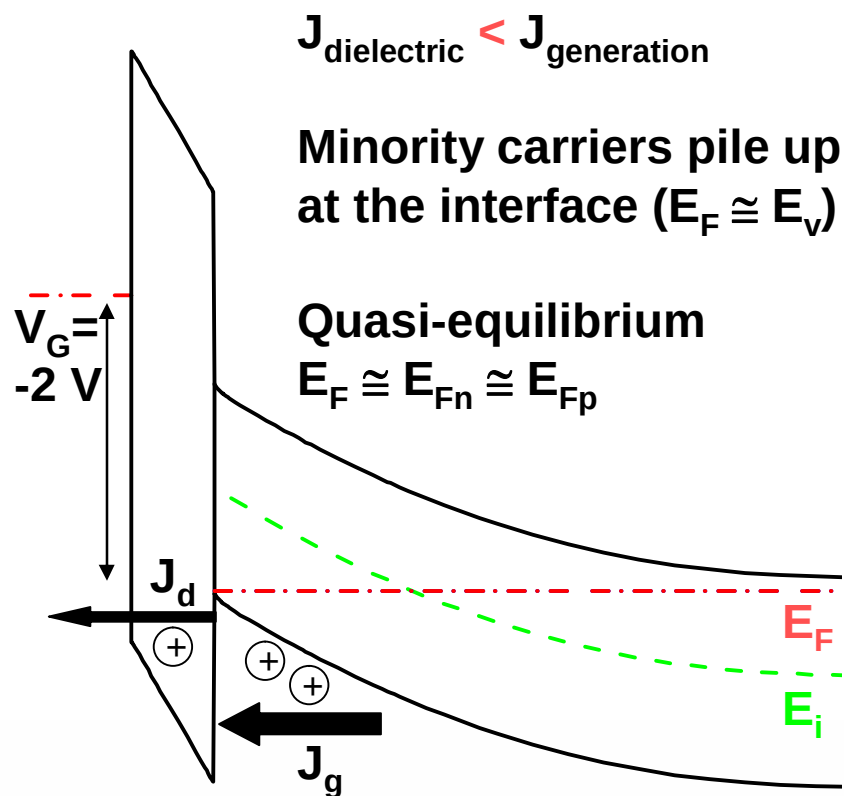
- *Start with a wide range of materials from InAs to GaAs*
 $0.35 \text{ eV} \leq \text{bandgap} \leq 1.42 \text{ eV}$: inversion vs. deep depletion condition
- *Subsequently focus on InAs: dispersion due to traps, selection of measurement techniques, parameters, and trap quantification model*

Admittance-Voltage: Flow Chart



Steady-State Criterion for Inversion (n-type)

INVERSION

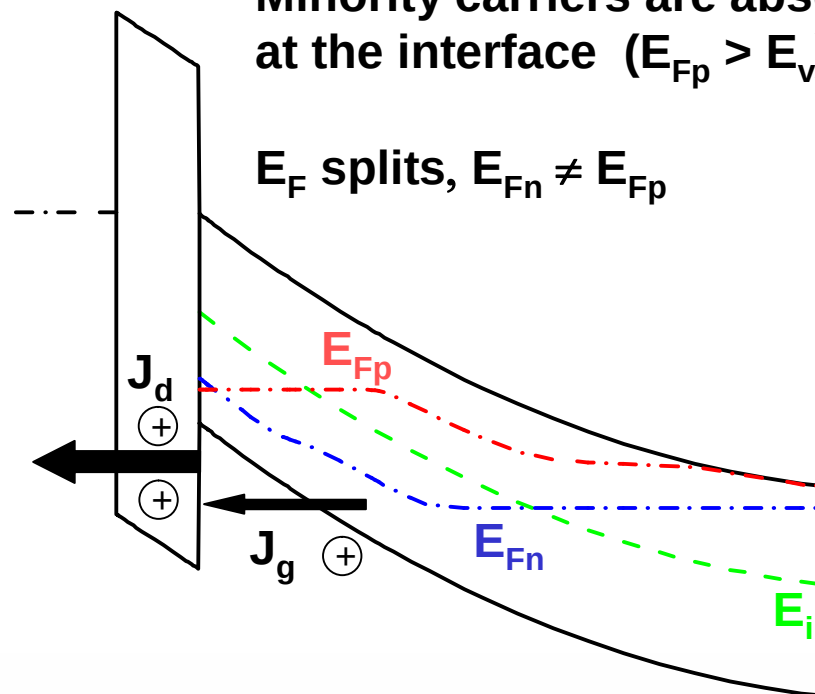


DEEP DEPLETION

$$J_{\text{dielectric}} > J_{\text{generation}}$$

Minority carriers are absent at the interface ($E_{Fp} > E_v$)

E_F splits, $E_{Fn} \neq E_{Fp}$

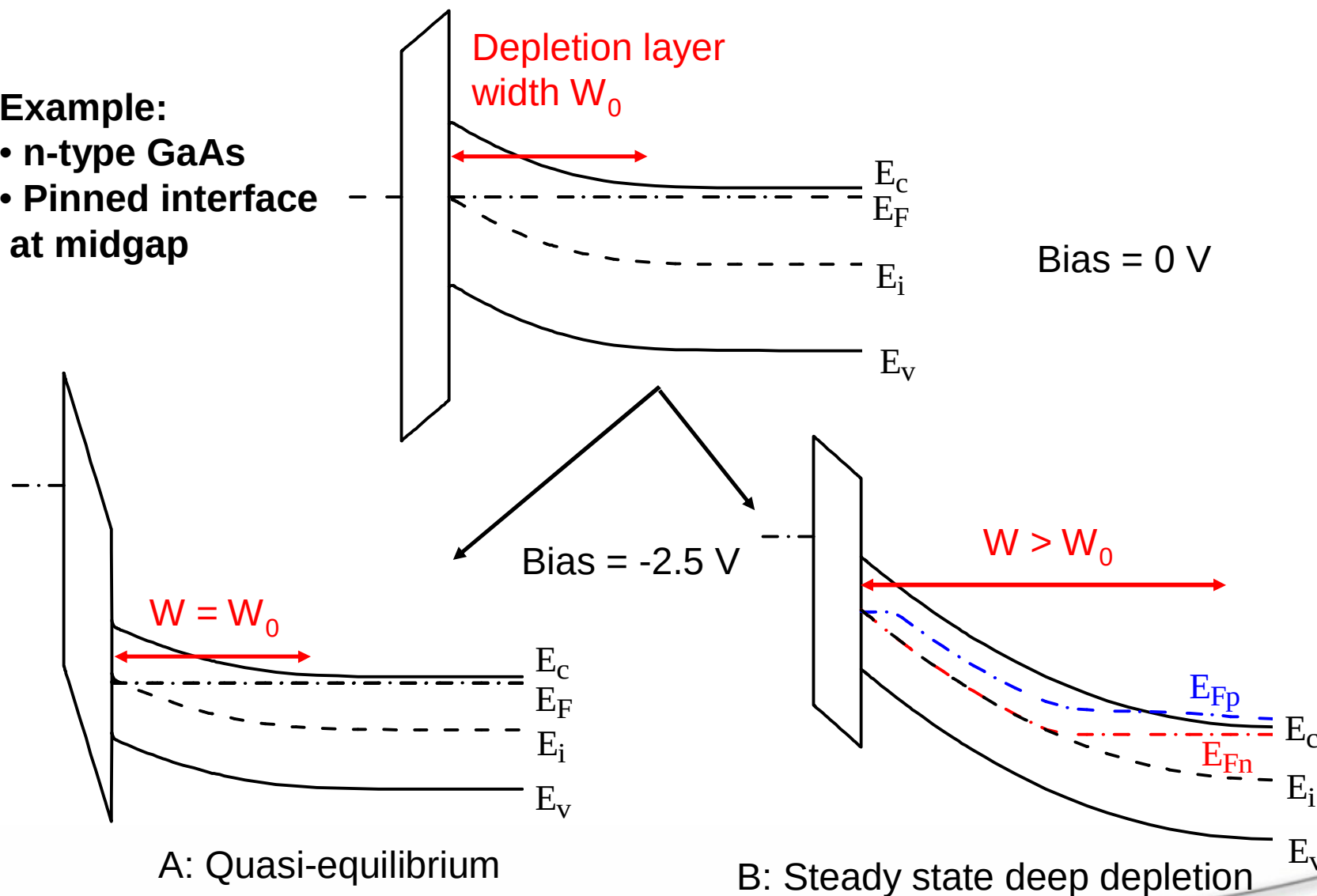


JAP, vol. 81, pp. 7647 (1997)

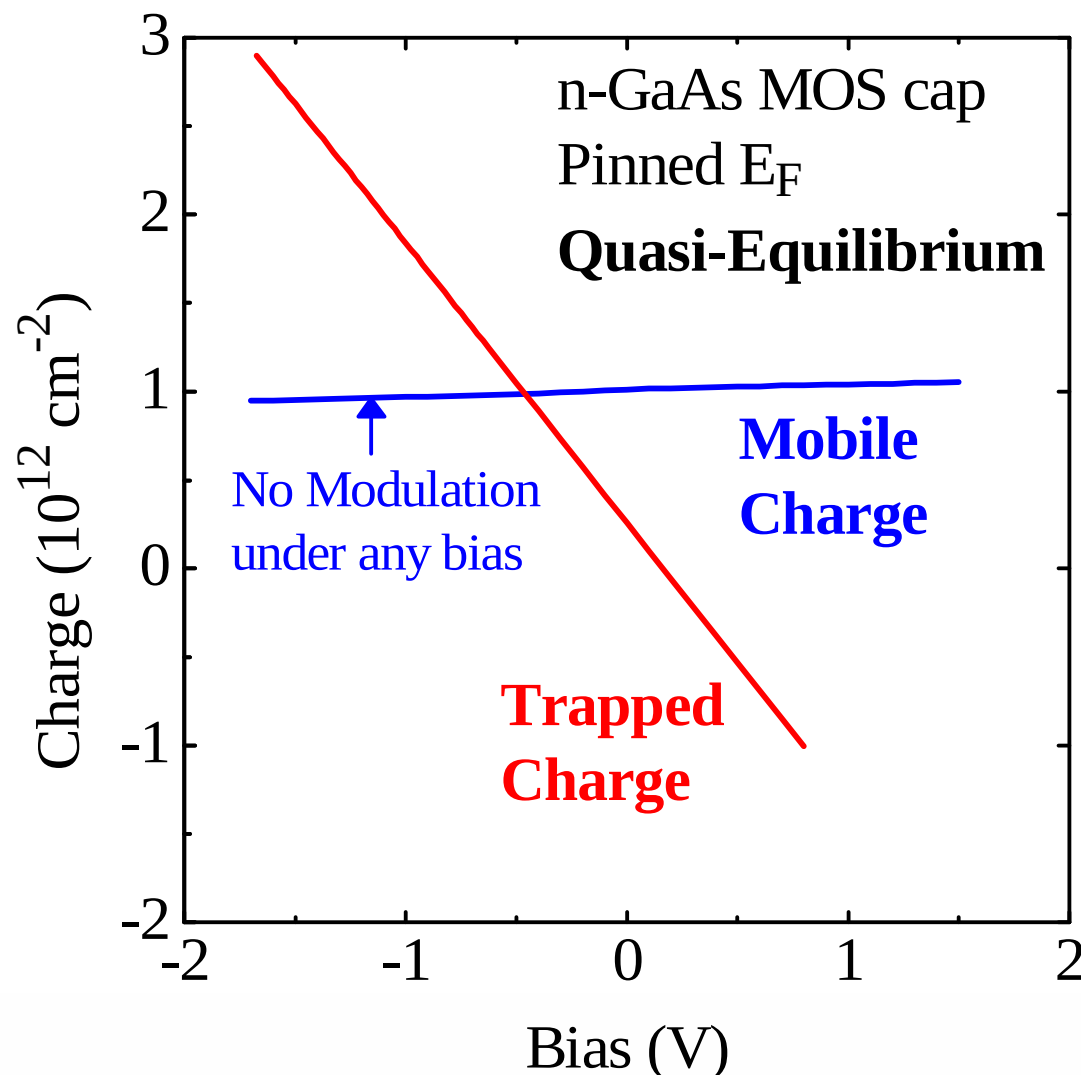
Quasi-Equilibrium vs. Steady-State Deep Depletion

Example:

- n-type GaAs
- Pinned interface at midgap



A: Quasi-Equilibrium and Pinned Fermi Level



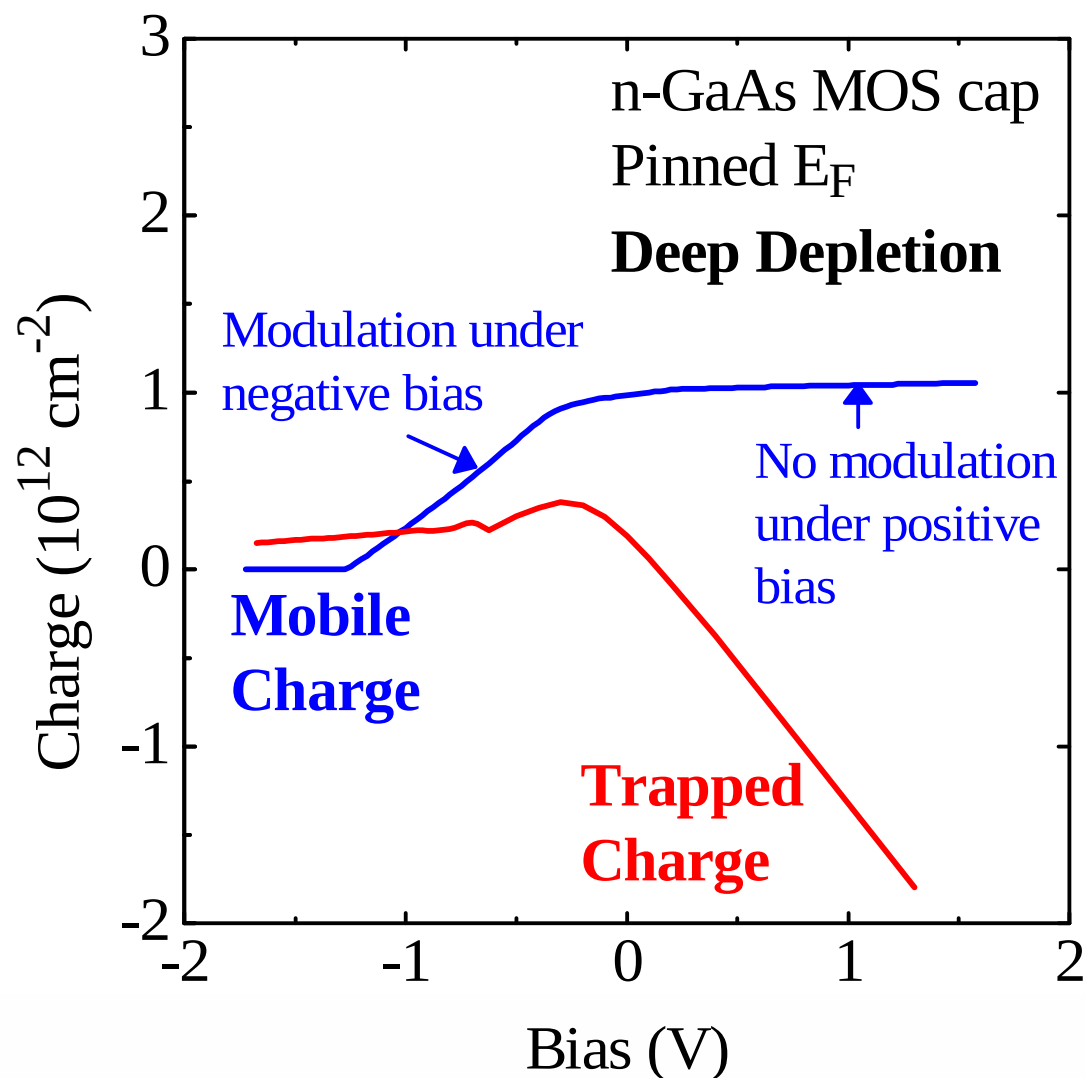
Quasi-equilibrium:

$\Rightarrow$ Mobile Charge **can not** be modulated

$\Rightarrow D_{it}$ **is** detectable at the capacitor's terminals

MOSFETs **do not** work

B: Deep Depletion and Pinned Fermi Level

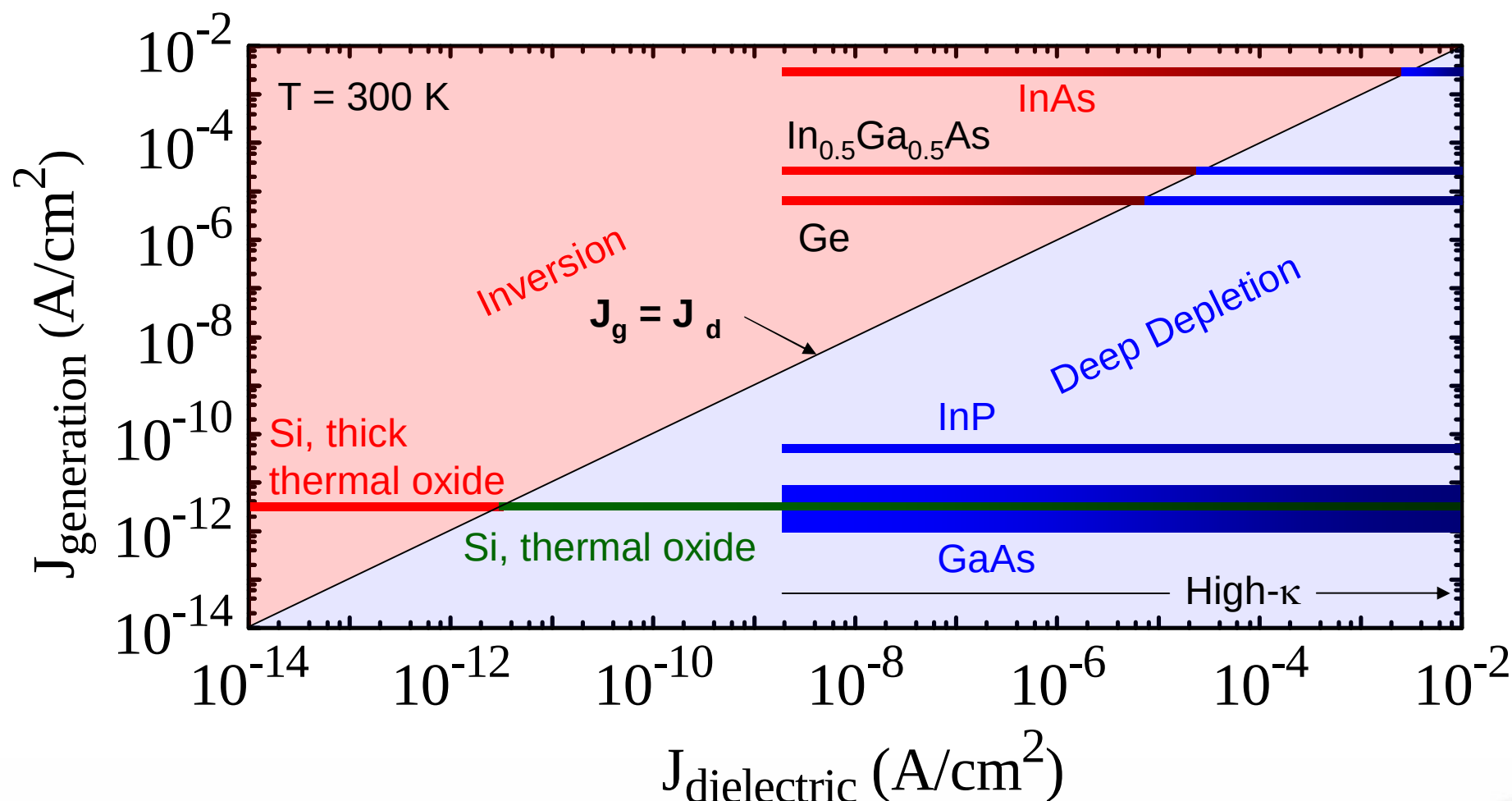


Positive Bias:
 $\Rightarrow$ No modulation of mobile charge possible (Enhancement-mode MOSFETs **do not** work)

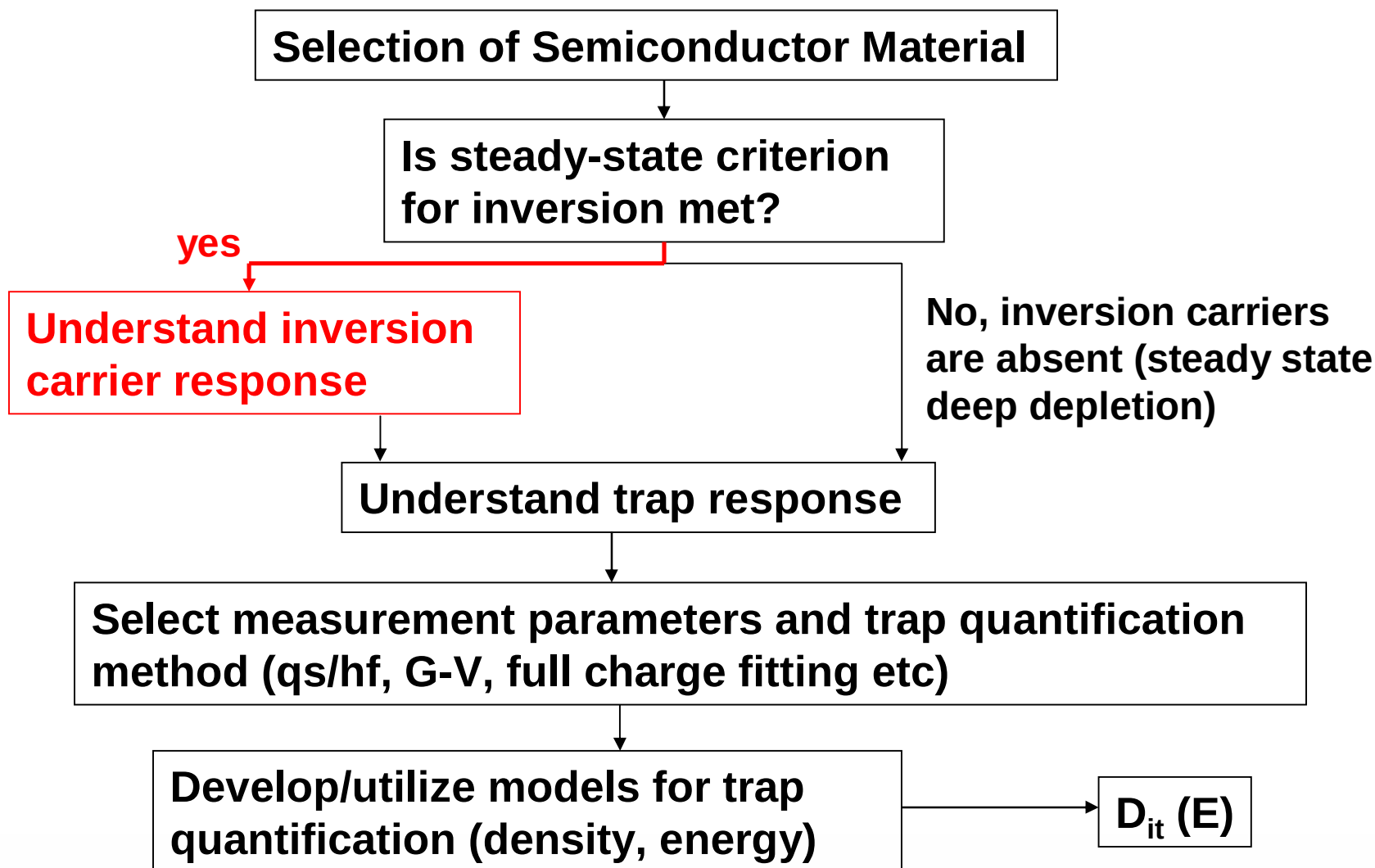
Deep Depletion:
 $\Rightarrow$ Mobile Charge **can be** modulated and depleted (Depletion-mode MOSFETs **can** work)

$\Rightarrow D_{it}$ **is not** detectable at the capacitor's terminals

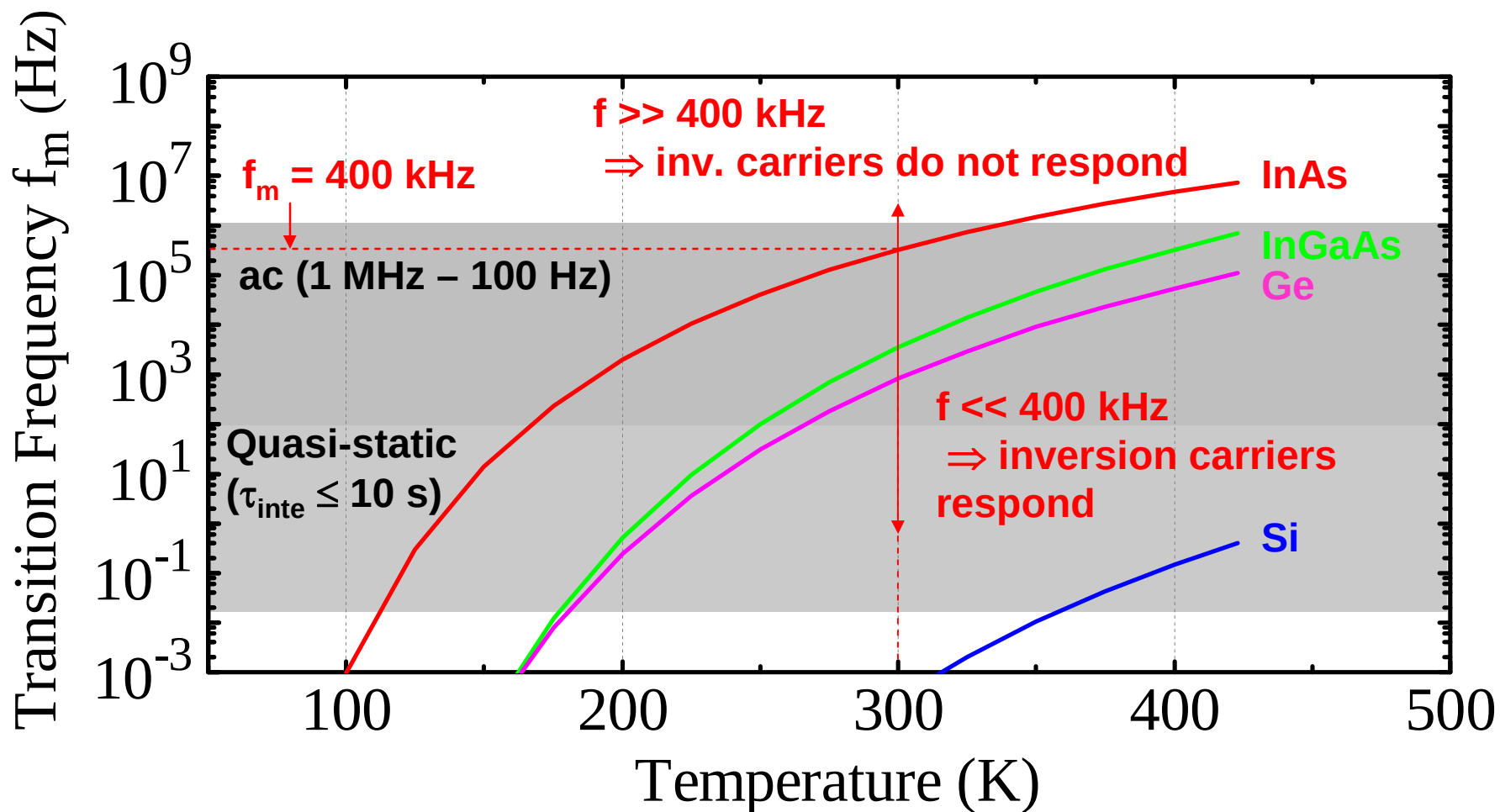
Steady-State Criterion for Inversion (300 K)



Admittance-Voltage: Flow Chart



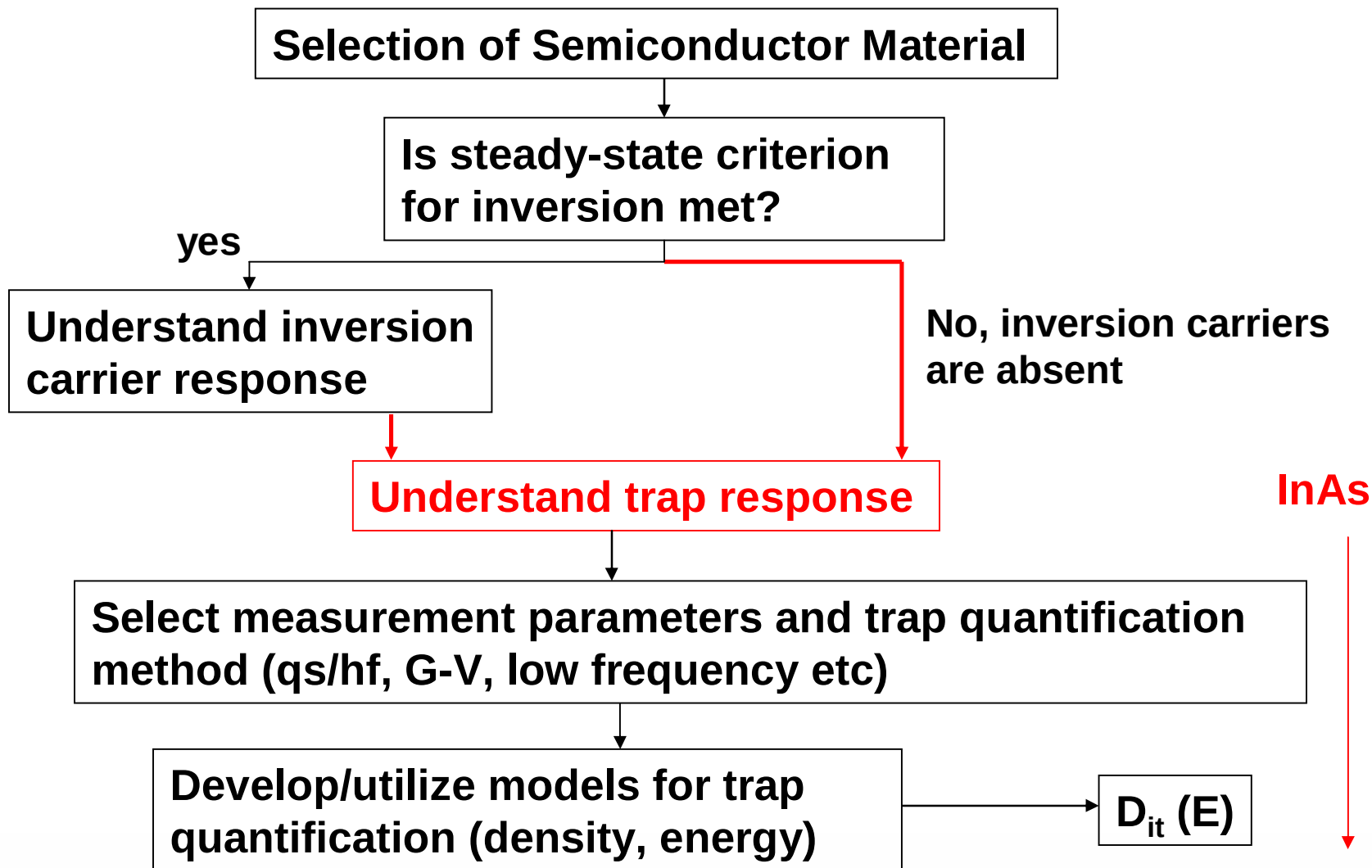
Inversion Carrier Response (f_m)



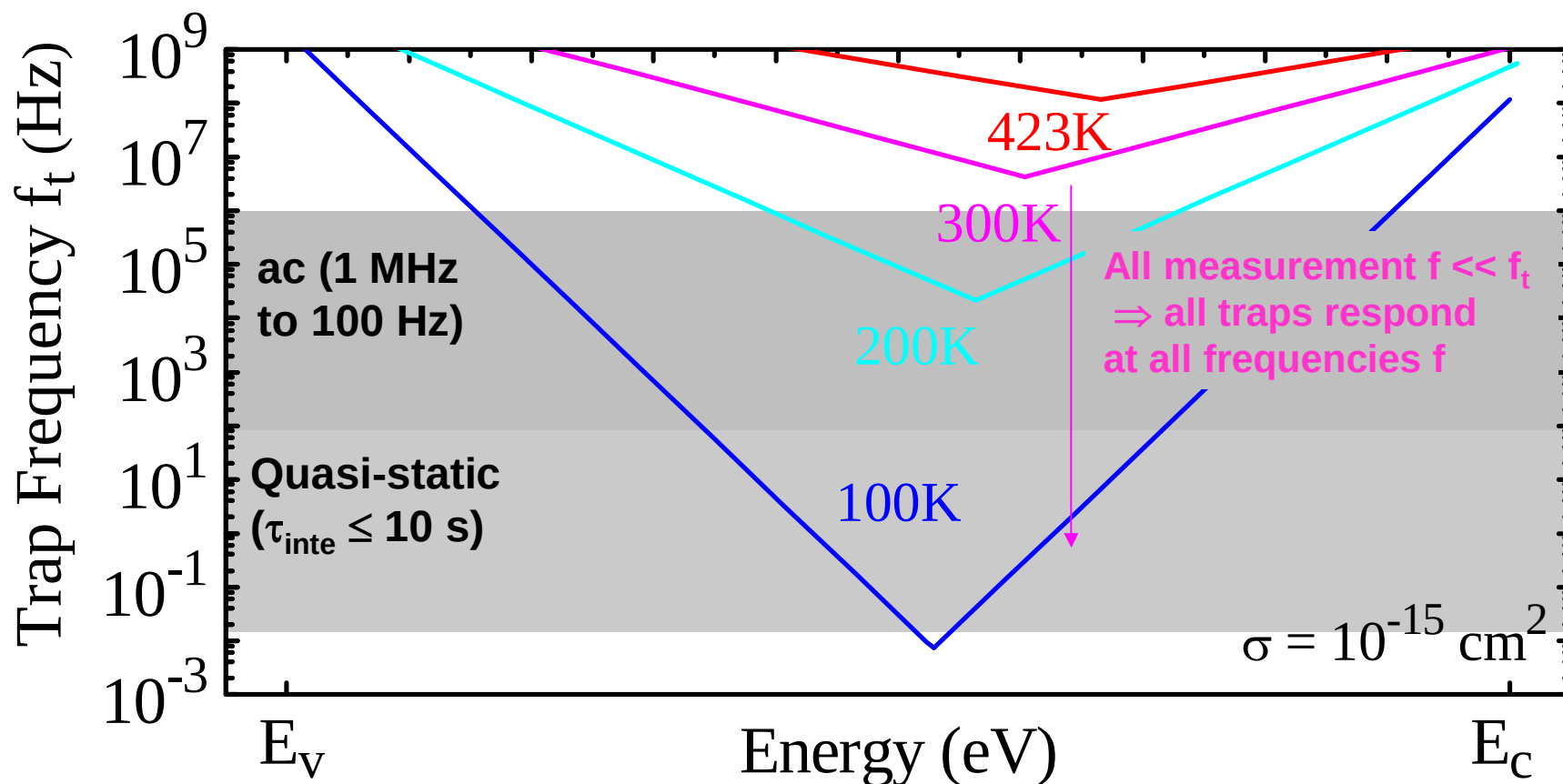
Example: InAs at 300 K

Parameters: bulk lifetime, doping density

Admittance-Voltage: Flow Chart

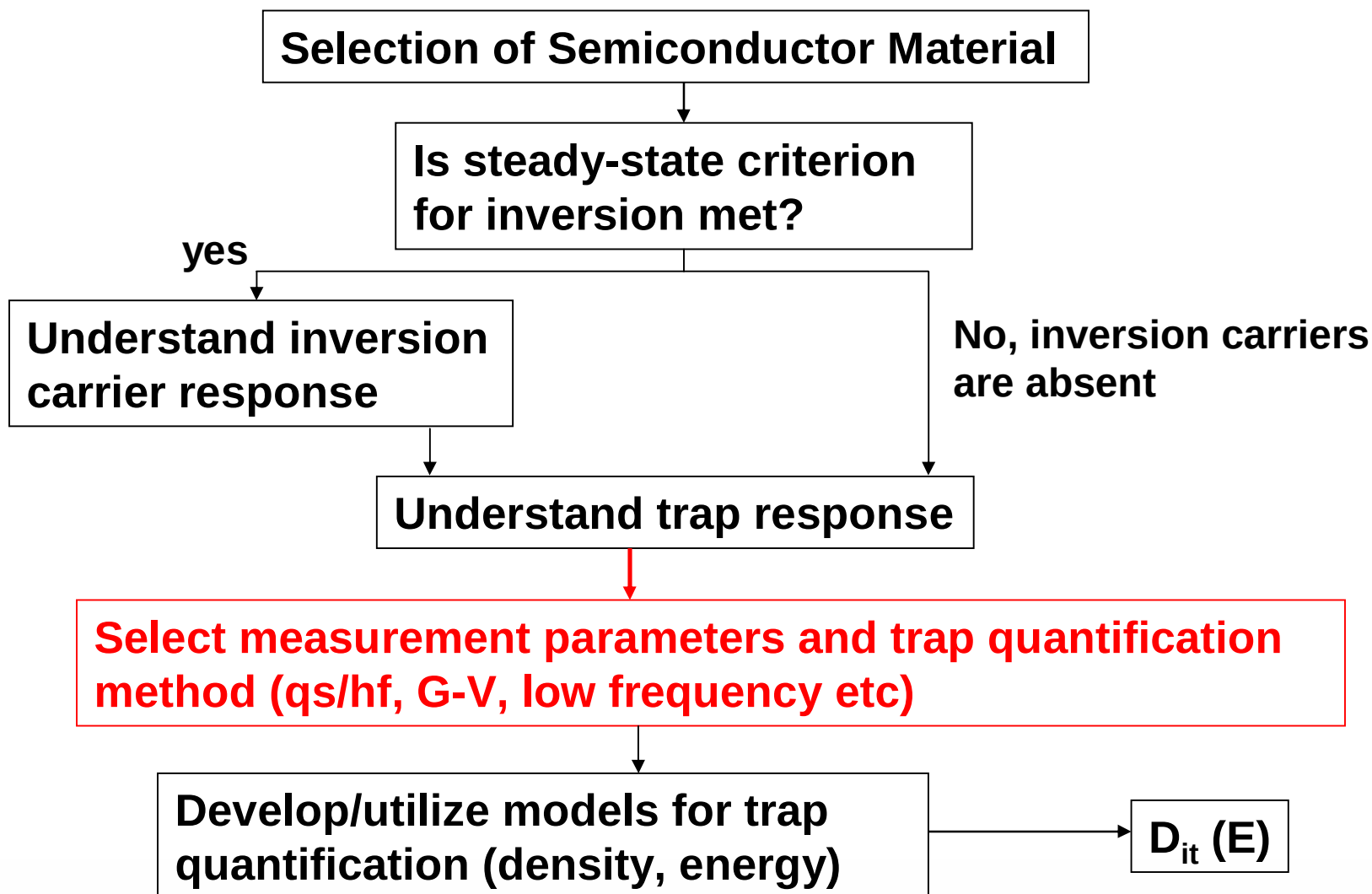


Trap Response (InAs)



Example: 300 K

Admittance-Voltage: Flow Chart



Measurement Parameters and Trap Quantification Method (InAs)

Findings (predictions):

- Inversion carriers respond for $f < 100$ kHz at 300K
 - All traps throughout the bandgap are expected to respond to $f < 100$ kHz at 300K ($\sigma = 10^{-15}$ cm²)
- ⇒ No frequency dispersion expected for $f < 100$ kHz

Selecting measurement parameters:

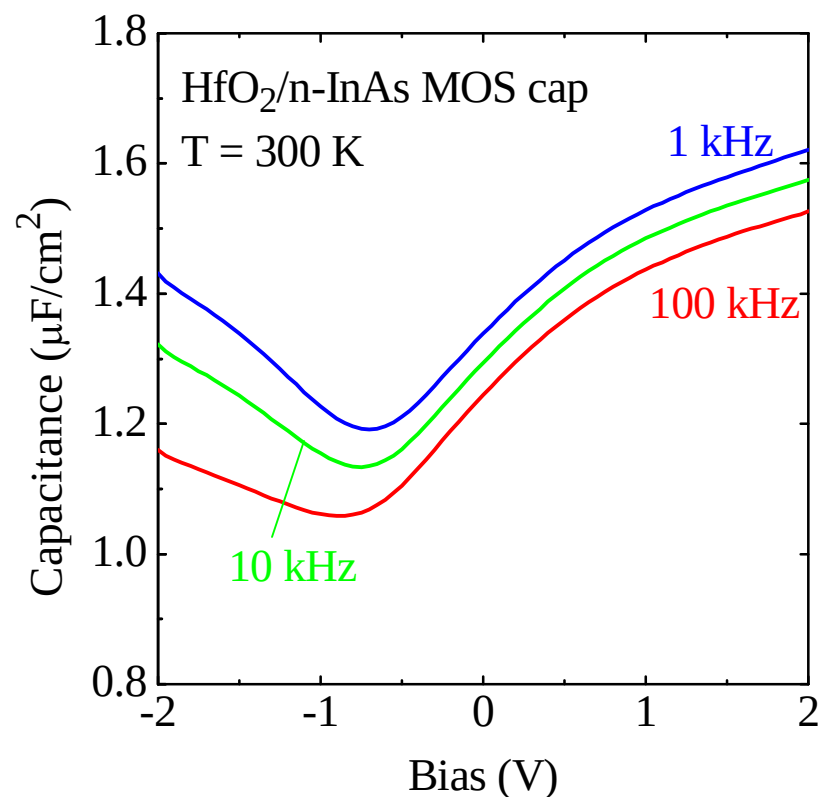
⇒ ac measurements (100 Hz – 100 kHz) at 300K

Selecting appropriate trap quantification method:

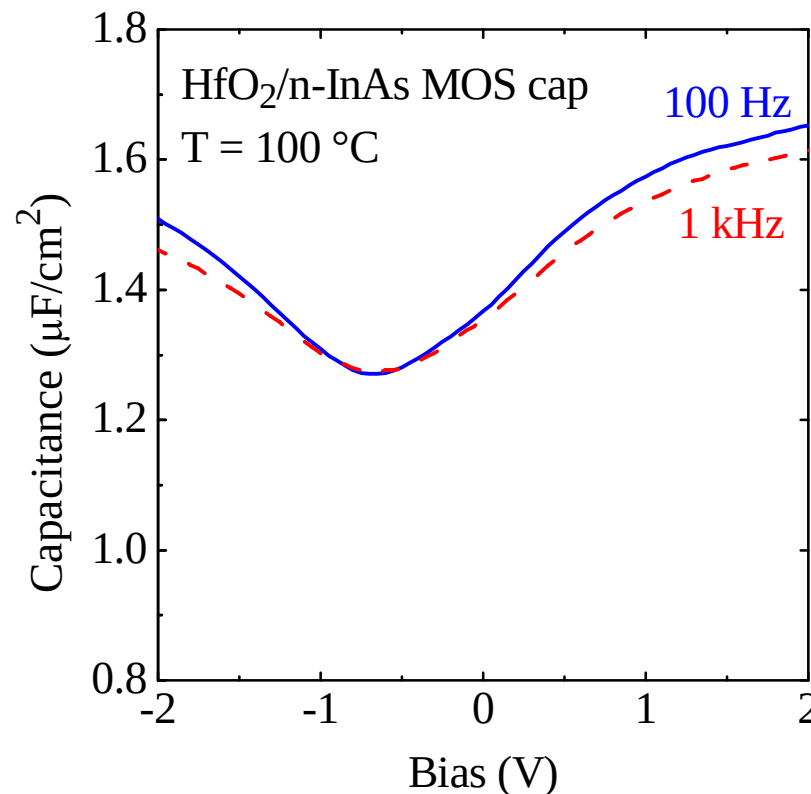
⇒ Low frequency C-V method (all traps must respond to dc sweep and ac signal) to extract D_{it} is most convenient

⇒ other methods such as I_f/hf , G-V, and Terman rely on partial trap response and don't appear to be as suitable (require to go to low temperature and/or much higher frequency)

Capacitance-Voltage measurements (InAs)



Frequency dispersion is experimentally observed at 300 K and below $f = 100$ kHz
contrary to expectations



C-V curves essentially collapse onto each other at elevated T and low $f \Rightarrow$ these curves are suitable for the low frequency method

Reconcile Prediction with Reality (InAs)

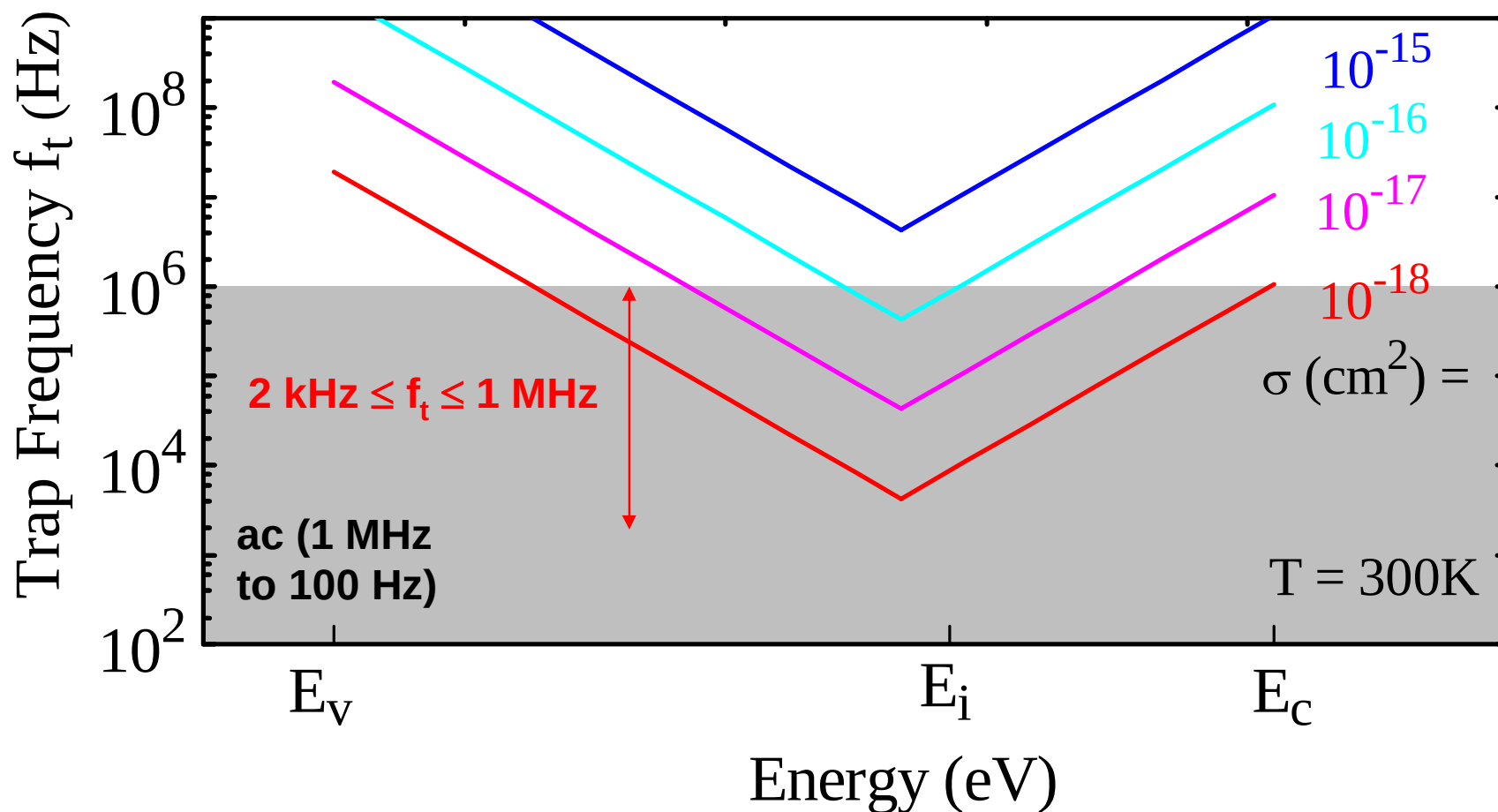
Prediction: No frequency dispersion for $f < 100$ kHz

Measurement: Dispersive between 1 MHz and 1 kHz, curves essentially collapse onto each other only for $f < 1$ kHz at elevated temperature

Prediction can be reconciled with reality assuming small (effective) capture cross section ($\leq 10^{-18}$ cm²)

- ⇒ this points toward border traps or oxide bulk traps
- ⇒ abruptness of interface is likely major issues

Effect of Capture Cross Section σ on Trap Response (InAs) TSMC Property



Experimentally observed frequency dispersion can be explained by small capture cross sections of e.g. $\sigma = 10^{-18} \text{ cm}^2$

Reconcile Prediction with Reality (InAs)

Prediction: No frequency dispersion for $f < 100$ kHz

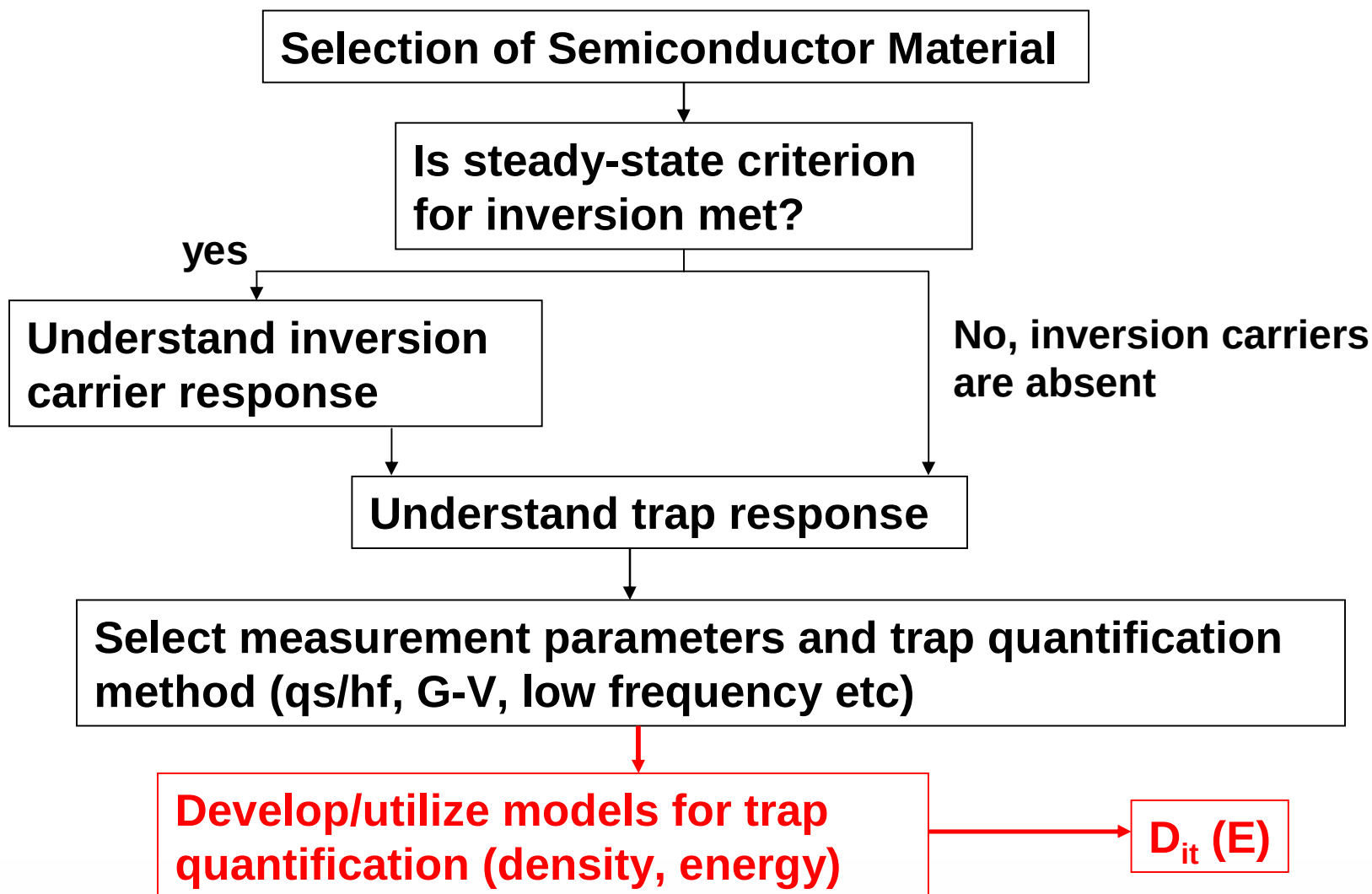
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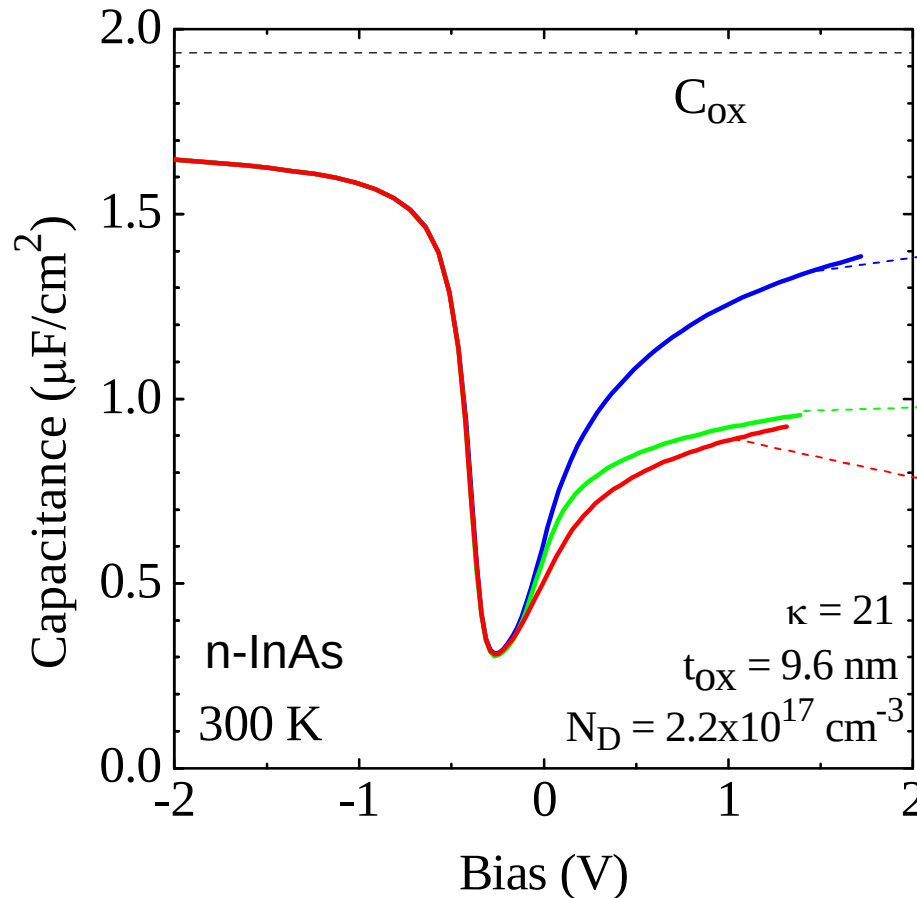
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Next: Low frequency method to determine D_{it} distribution from 100 Hz, 100 °C C-V data

Admittance-Voltage: Flow Chart



D_{it} Quantification: Low Frequency Method (InAs)



Modeling

Classic, nonparabolic (NP)

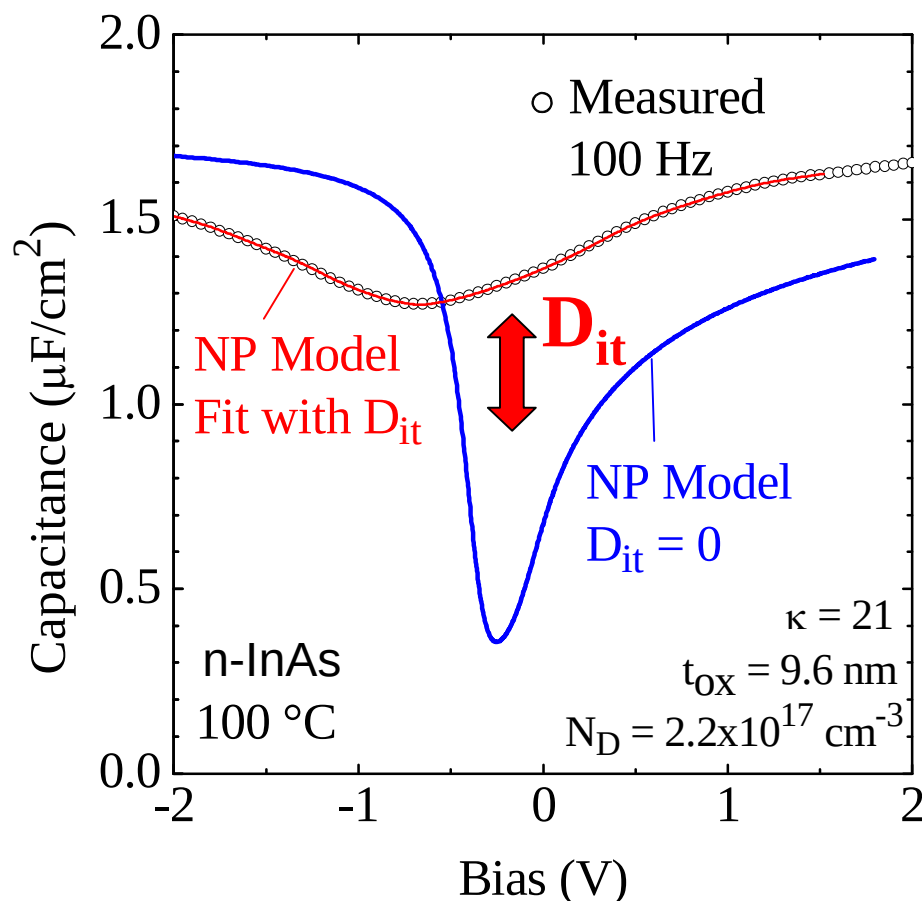
Classic, parabolic

Quantum mechanical,
parabolic

Low effective density of states (DOS) in conduction band ($8.7 \times 10^{16} \text{ cm}^{-3}$) causes asymmetry of C-V curves

- Model needs to include Fermi-Dirac statistics, nonparabolicity, quantization
⇒ accurate solution requires tight binding or k·p
- NP model is within 5-10% for bulk InAs (E. Lind et al., APL vol. 96, 2010)

D_{it} Quantification: Low Frequency Method (InAs)



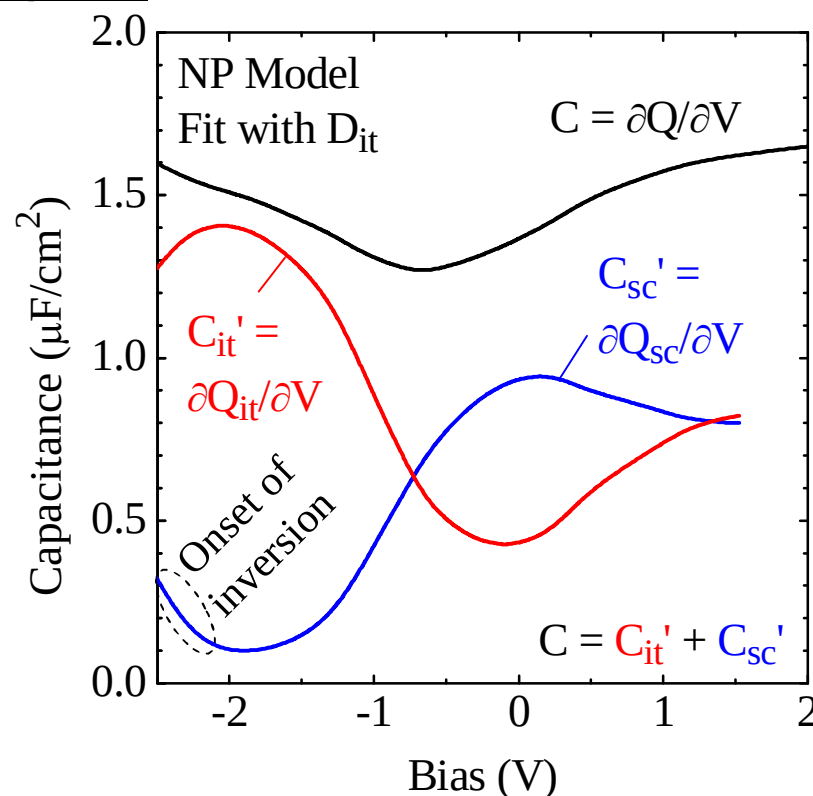
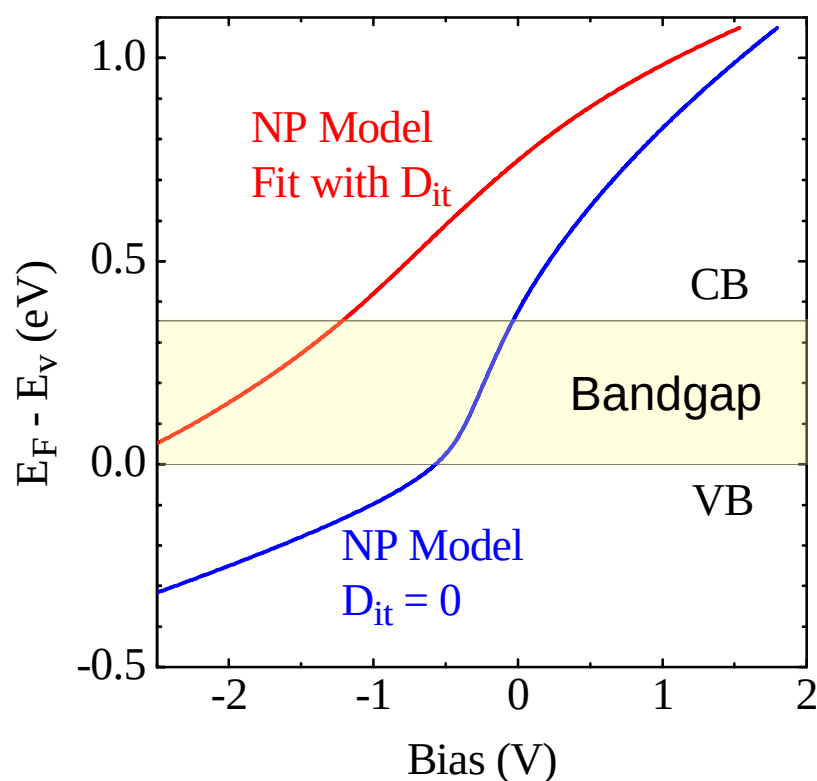
Simulations are fitted to measured MOS cap results using numerical technique

$\Rightarrow$ **Excellent fit of modeled data to MOS cap results**

Discrepancy between ideal model ($D_{it} = 0$) and experimental capacitor data due to D_{it}

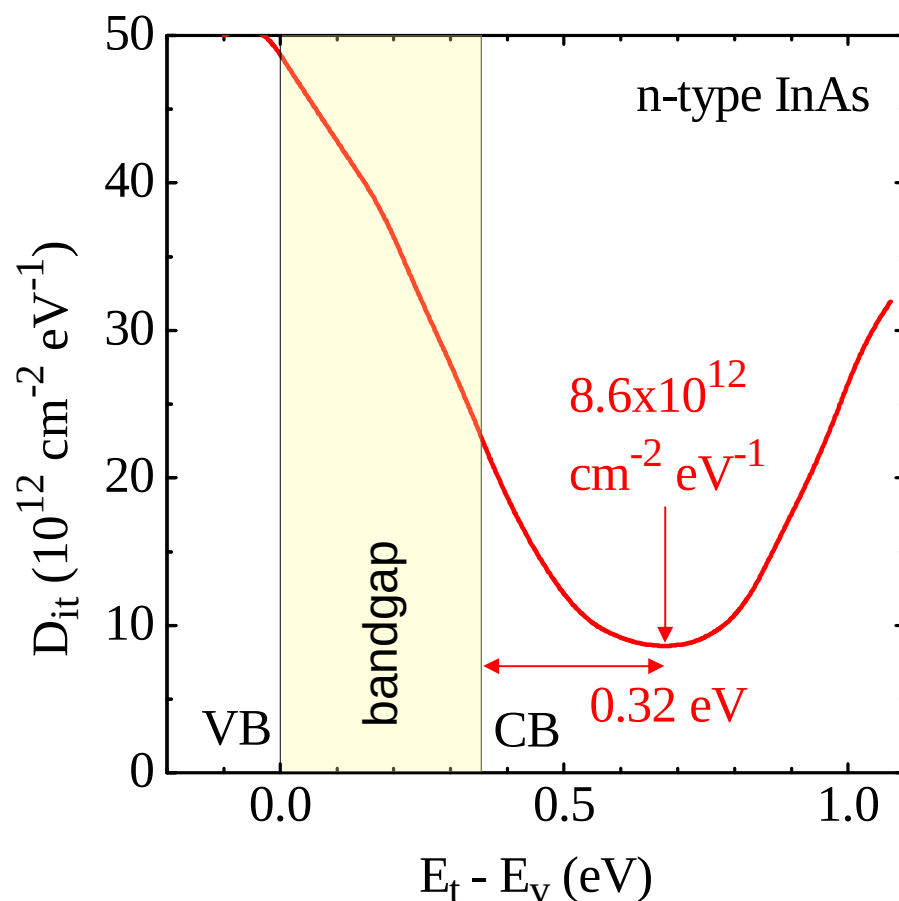
D_{it} Quantification: Low Frequency Method (InAs)

Insights



- Due to interface traps, Fermi level at interface does not move into VB
- No minority carrier contribution to gate capacitance down to -2V (upturn of total capacitance is due to D_{it})

D_{it} Quantification: Low Frequency Method (InAs)



Final result: D_{it} distribution

- D_{it} obtained throughout the InAs band gap and inside CB
- D_{it} minimum has density of $\cong 10^{13} \text{ cm}^{-2} \text{ eV}^{-1}$ at $\cong 0.3 \text{ eV}$ inside CB

Conclusions and Acknowledgement: D_{it} Quantification on InAs

D_{it} minimum $\cong 10^{13} \text{ cm}^{-2} \text{ eV}^{-1}$ and is located at 0.3 eV above conduction band minimum

⇒ Inadequate for devices

Virtually all frequency dispersion originates from traps

⇒ Many traps have very small ($\leq 10^{-18} \text{ cm}^2$) effective capture cross section

⇒ this points toward border traps or oxide bulk traps

⇒ abruptness of interface is likely major issues

Inversion carriers (holes) of noticeable density are absent

⇒ Inversion carrier response time becomes a moot point

Acknowledgement

- Georgios Vellianitis: C-V measurement
- Gerben Doornbos: C-V and D_{it} modeling

Further Reading

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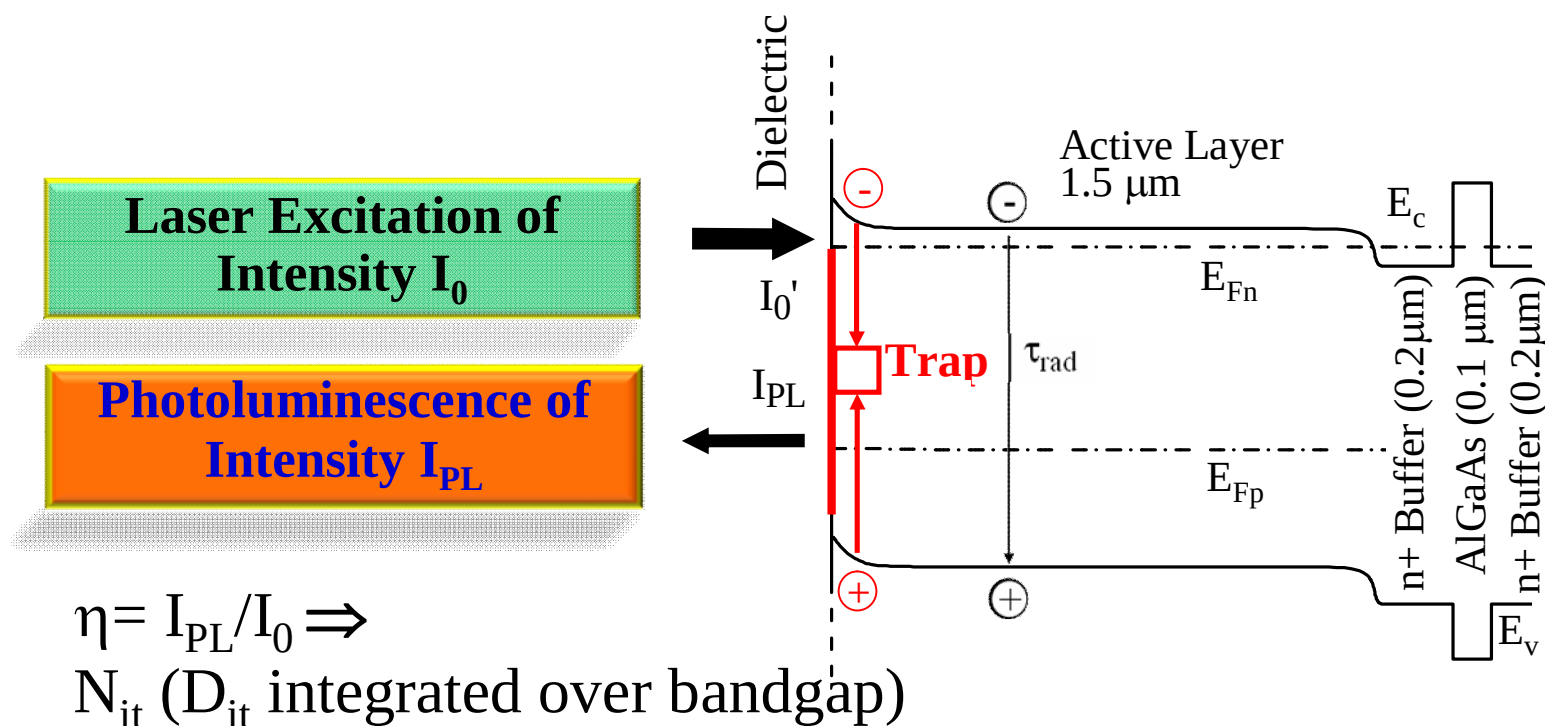
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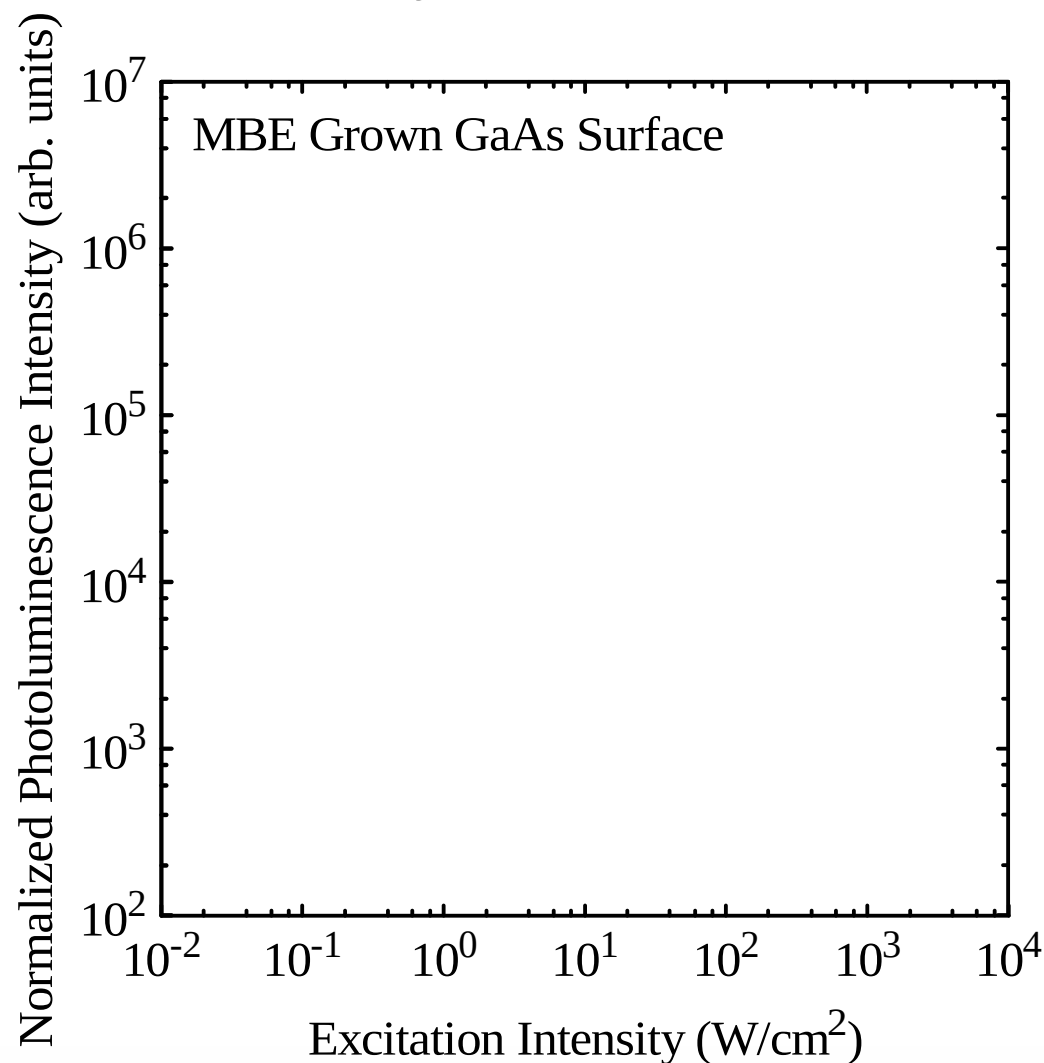
Trap (D_{it}) Analysis Using Recombination



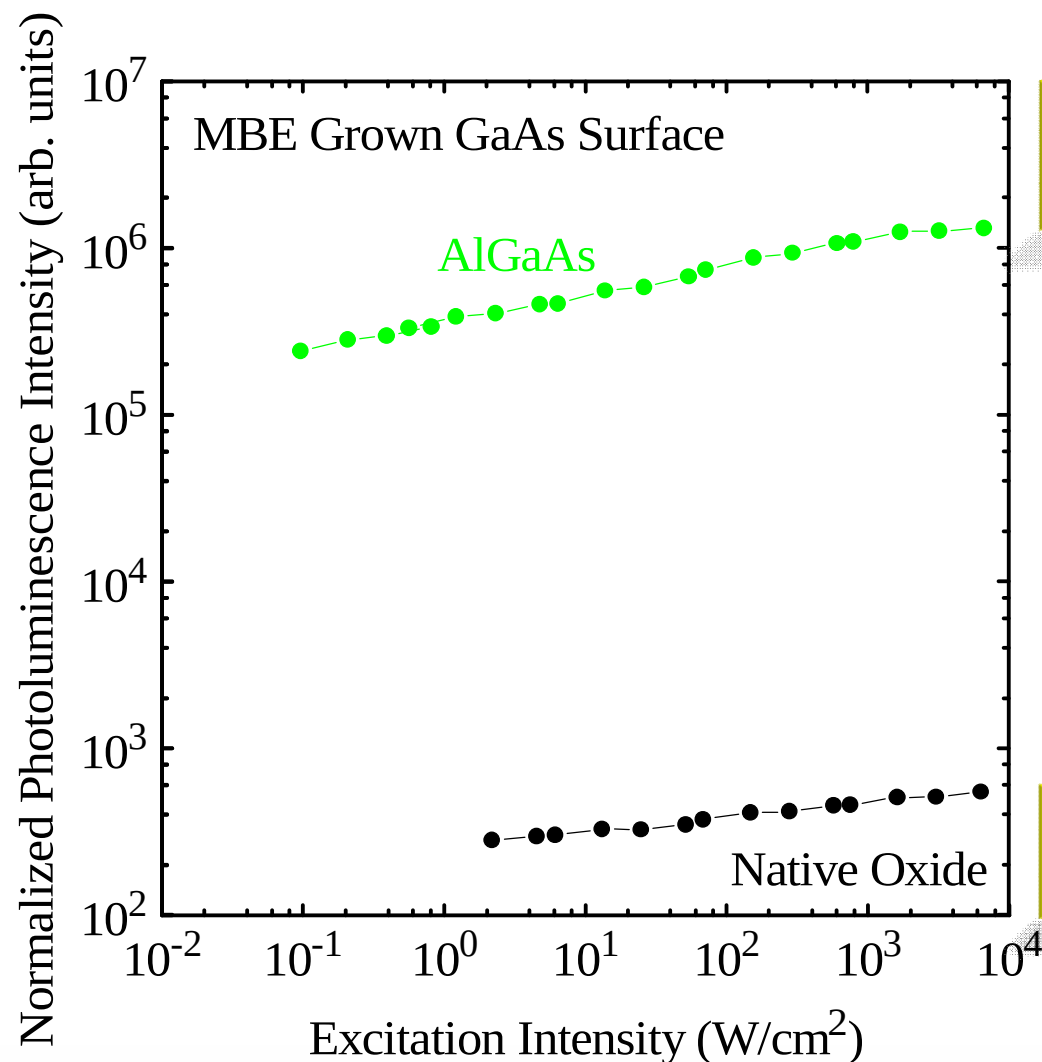
$$\text{Carrier Lifetime } \tau_{n/p} = 1/(\sigma v_{th} N_{it}) \neq f(E_T, n_i)$$

Unique Tool to Screen Interfaces for Device Quality

Trap (D_{it}) Analysis Using Recombination



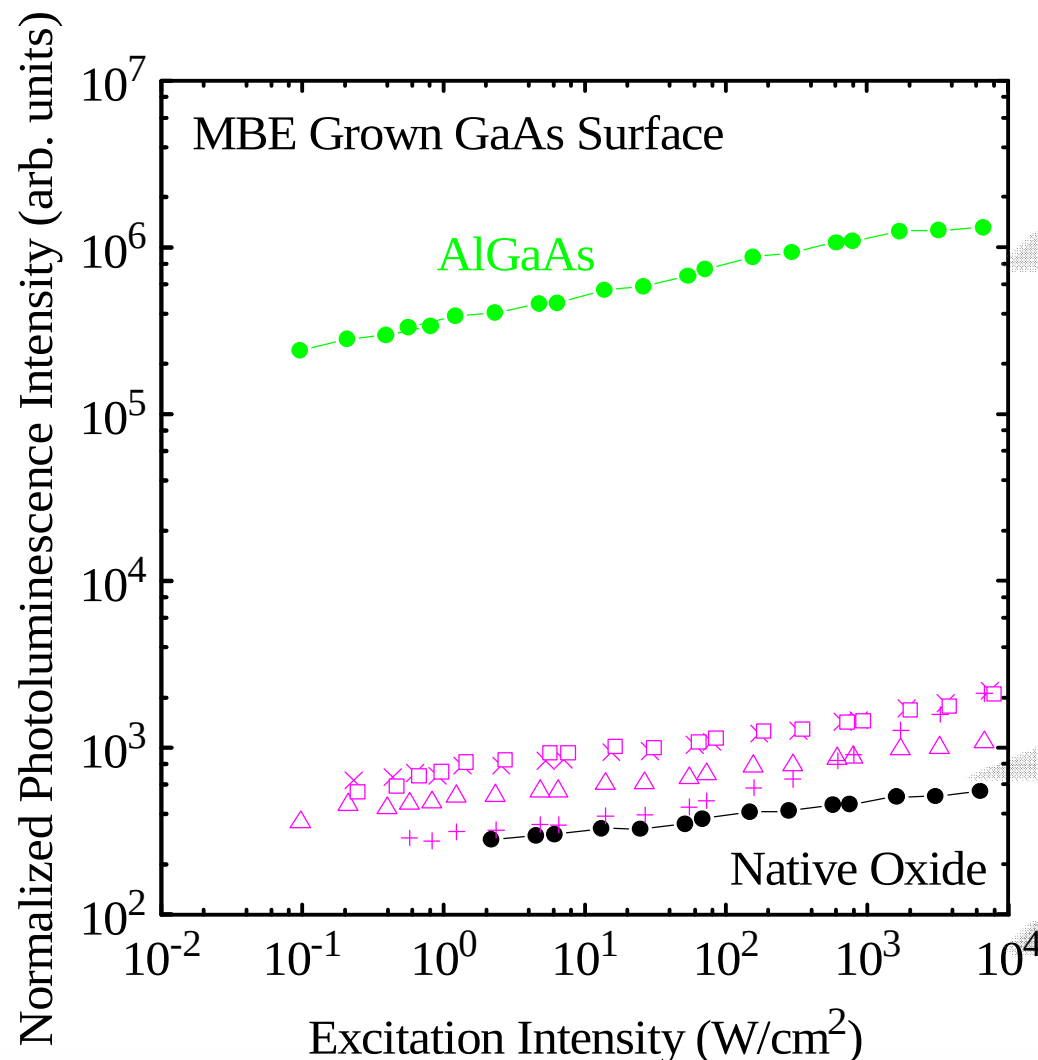
Trap (D_{it}) Analysis Using Recombination



Known Best Interface:
AlGaAs/GaAs

Known Worst Interface:
Native Oxide/GaAs

Trap (D_{it}) Analysis Using Recombination

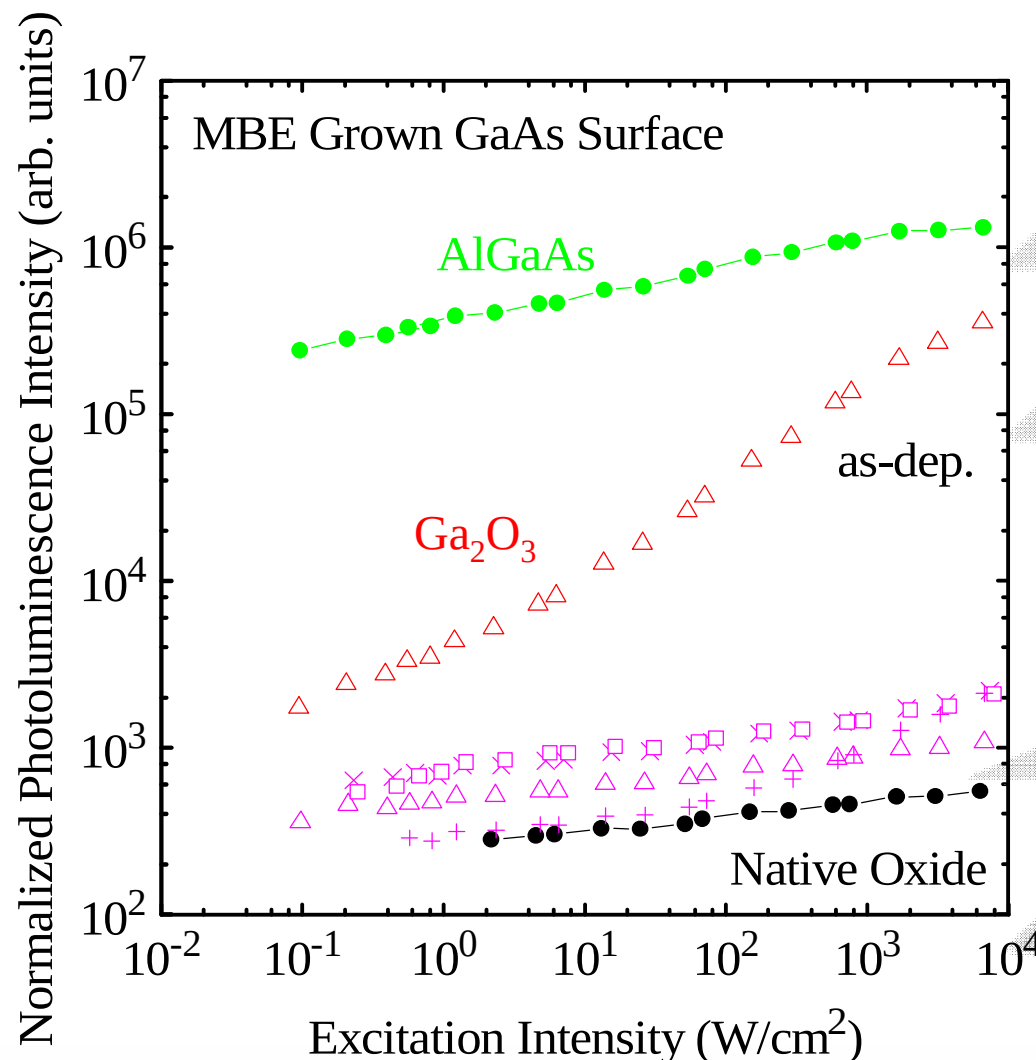


Known Best Interface:
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Oxides, Nitrides on GaAs:
Native Oxide Behavior

Known Worst Interface:
Native Oxide/GaAs

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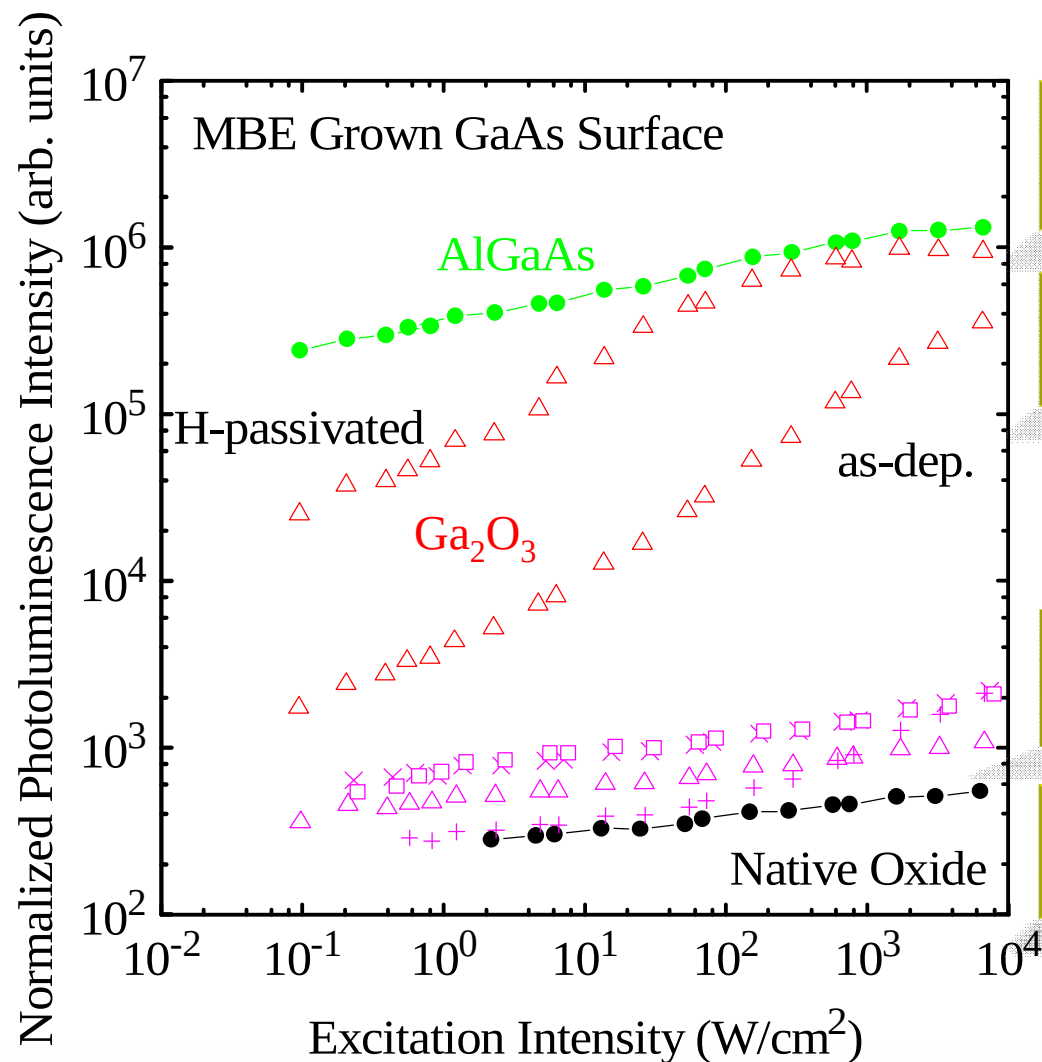
Known Best Interface:
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Device Quality Interface:
Ga₂O₃/GaAs

Oxides, Nitrides on GaAs:
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Known Worst Interface:
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Trap (D_{it}) Analysis Using Recombination



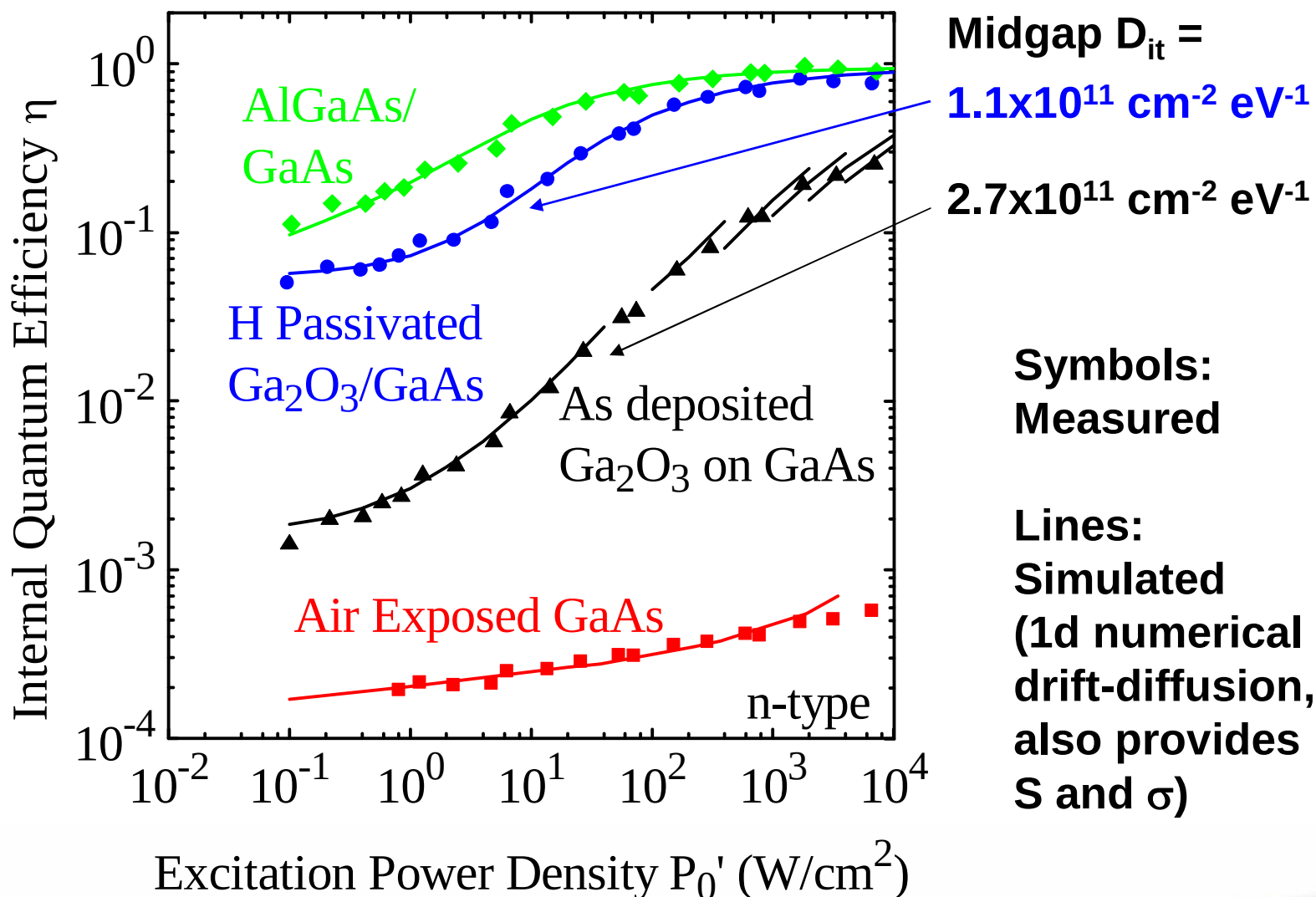
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Further Reading

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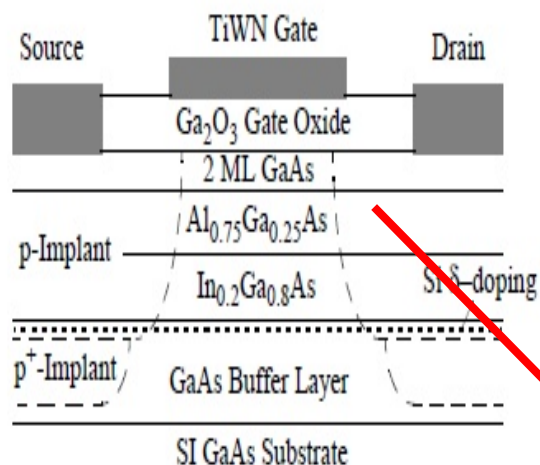
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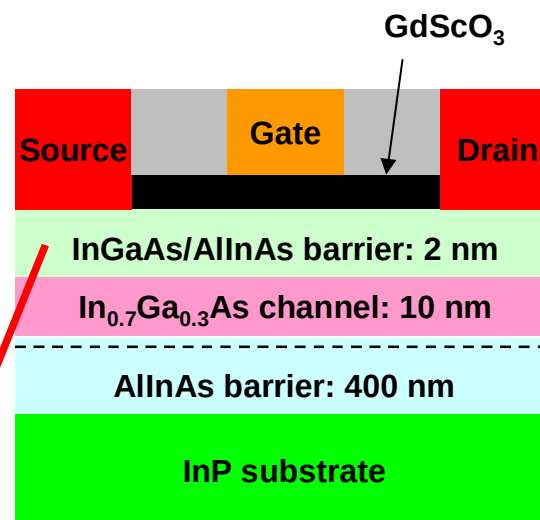
Quiz Solution

D_{it} and Heterostructures in Quantum Well MOSFETs (QWFET)

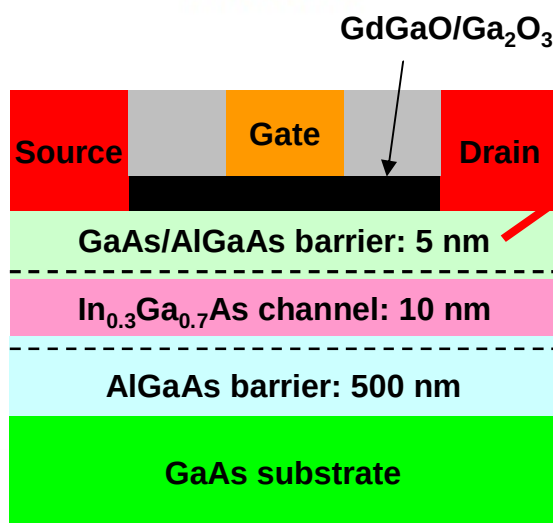


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pp. 498, pp.
1283

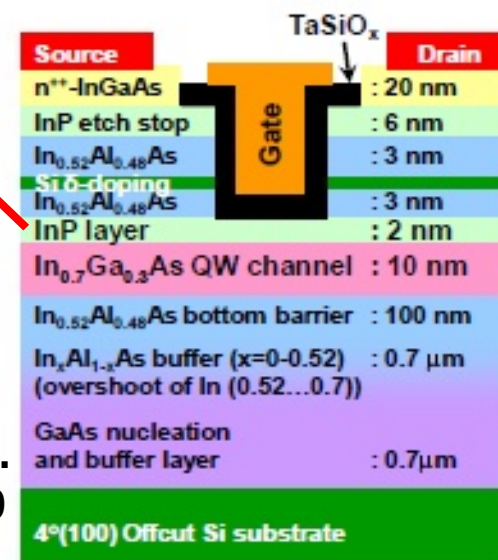


Heterostructure Barrier Layer



Hill et al.,
IEEE EDL
vol. 28, 2007,
pp. 1080

Radosavljevic et al.
IEDM 2009, pp. 319



Heterostructure Barrier Layer: Introduction

Alter surface chemistry and interface bonding

- Lower D_{it}
- More favorable D_{it} distribution

Shift a **given** D_{it} vs E distribution with respect to channel charge

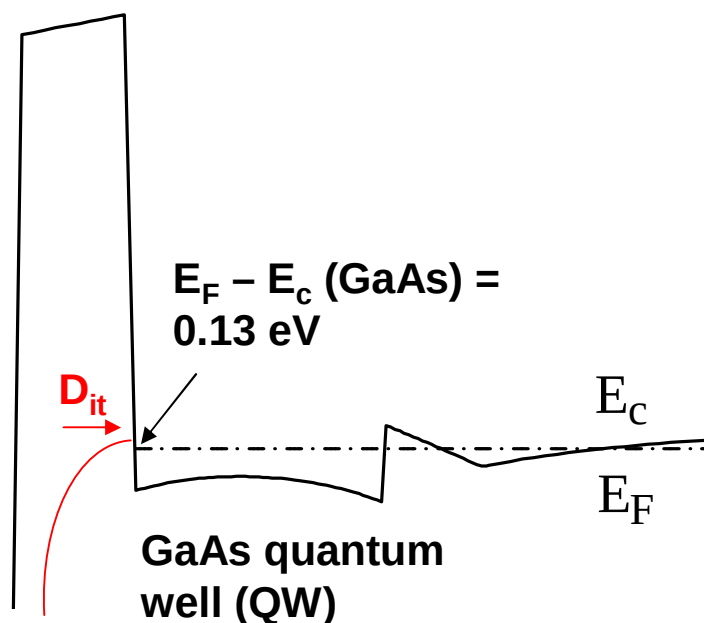
- Avoid branch of high/rising D_{it} , e.g. close to a band edge

Optimization considerations (**electrostatics only**)

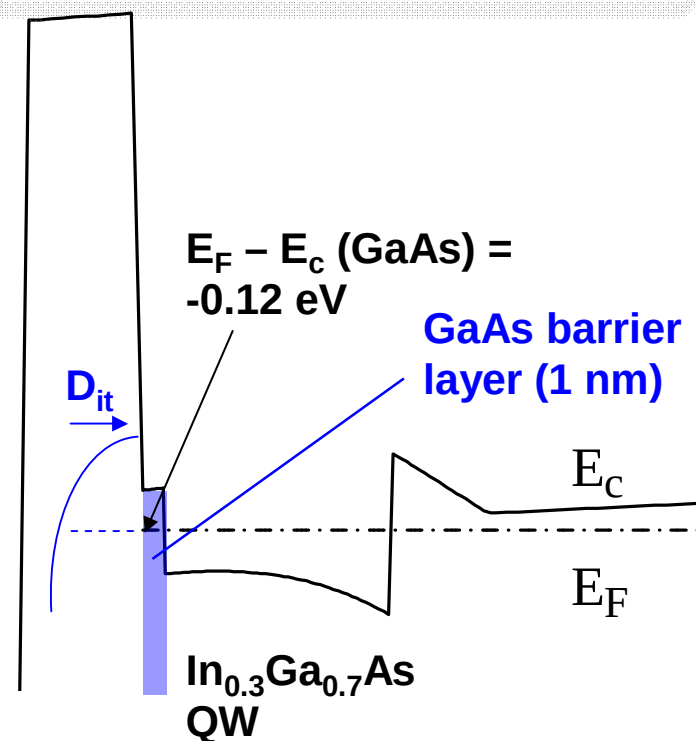
- EOT
- Effective D_{it}
- Leakage current

Shift of D_{it} vs E with respect to Channel Charge

No barrier layer

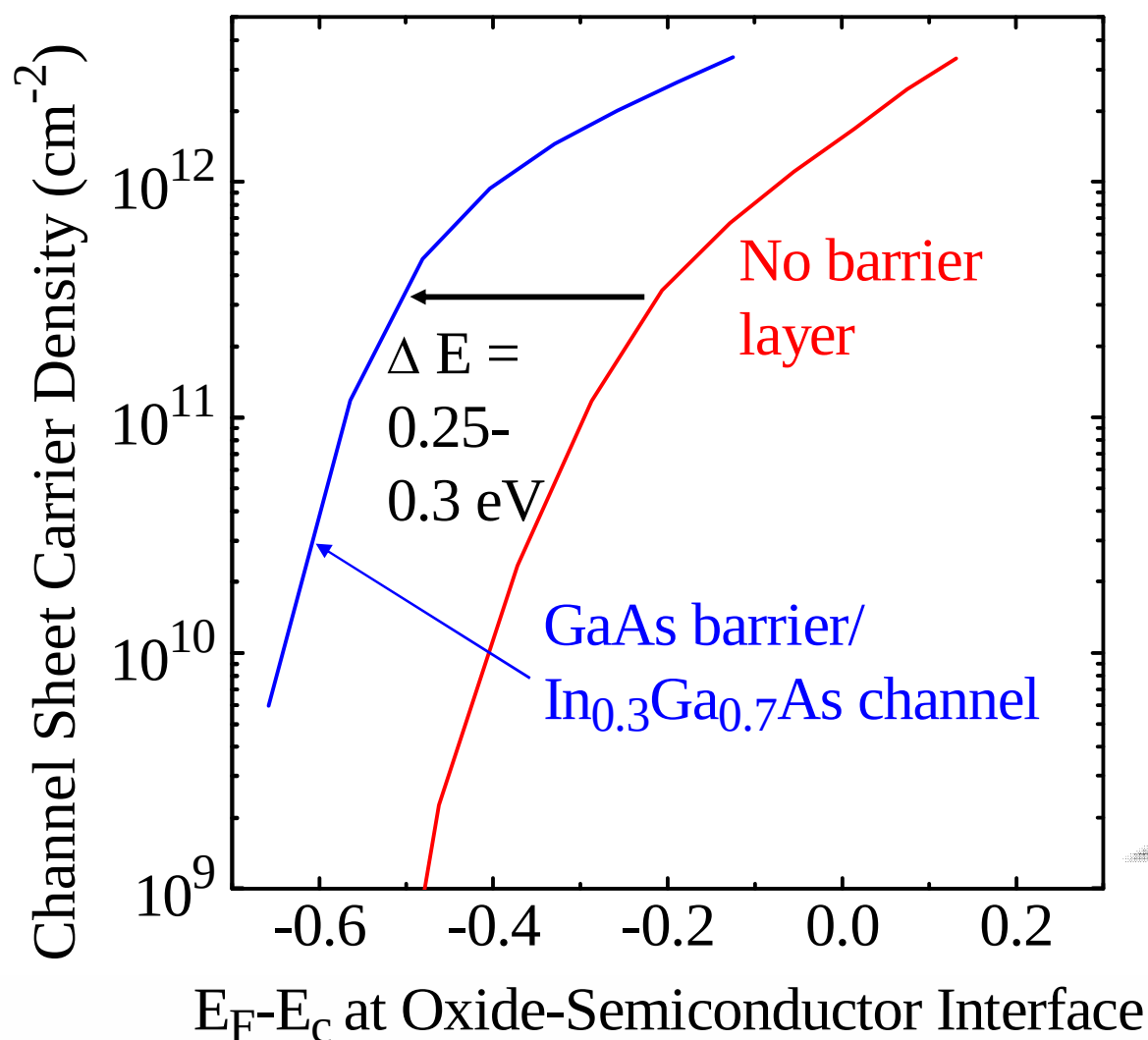


With Barrier Layer



Fermi level position at dielectric-GaAs interface differs by 0.25 eV
for identical QW sheet charge density: $n_s = 3.4 \times 10^{12} \text{ cm}^{-2}$

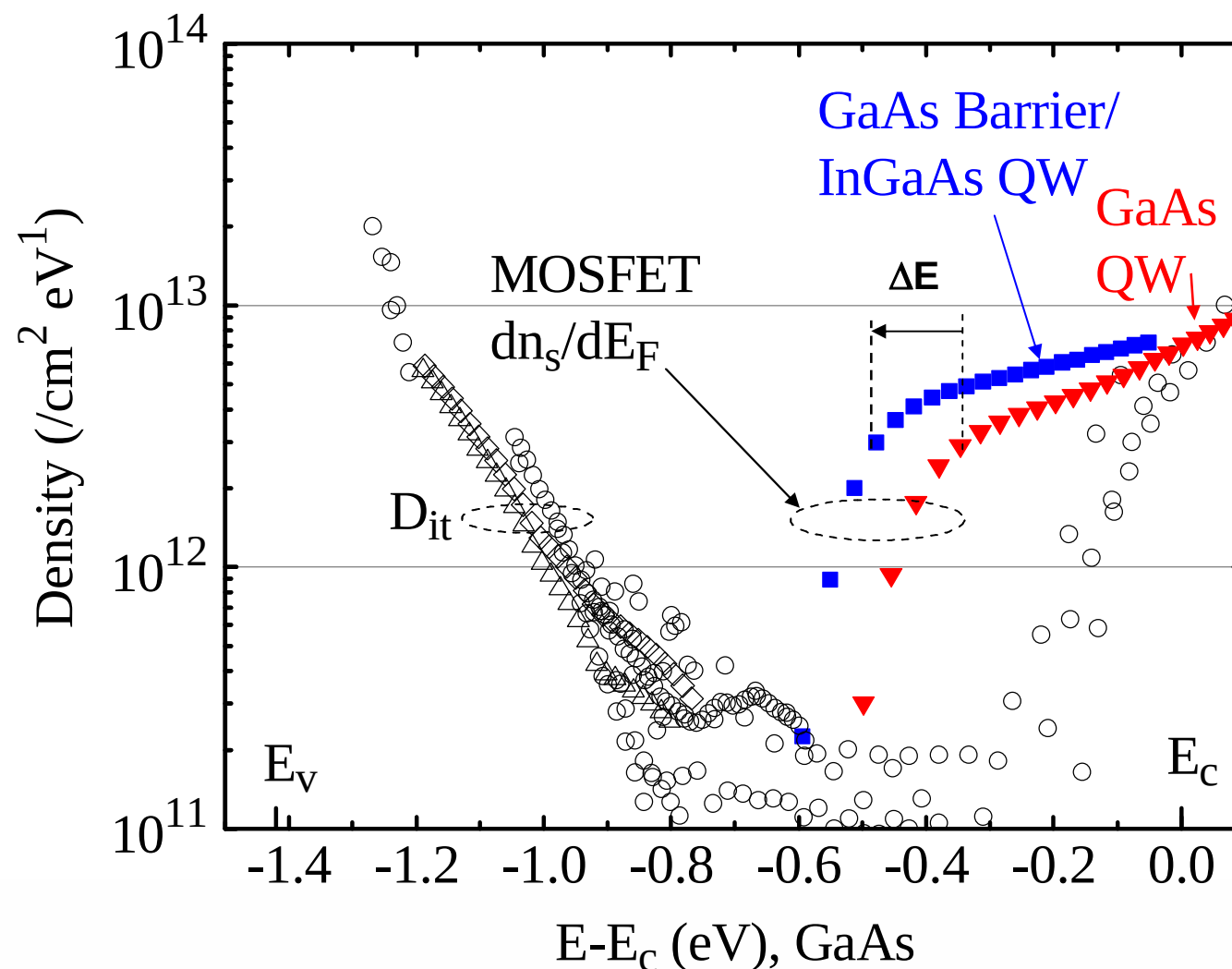
Shift of D_{it} vs E with respect to Channel Charge



D_{it} distribution shifted along energy axis by 0.25-0.3 eV for identical channel charge n_s

Model & Simulations

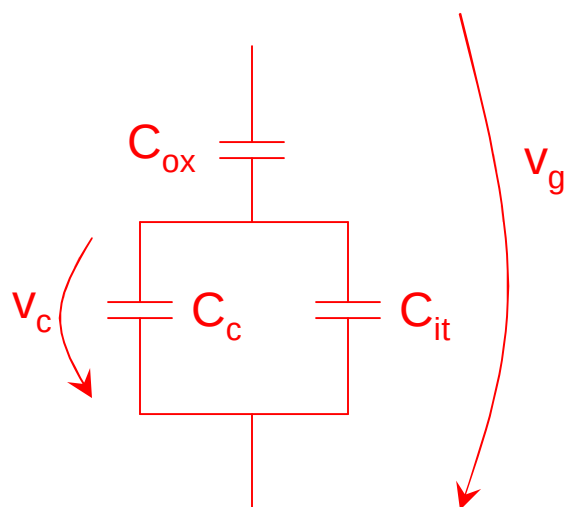
Shift of D_{it} vs E with respect to Channel Charge



Manufactured
Devices
(GaAs QWFET)

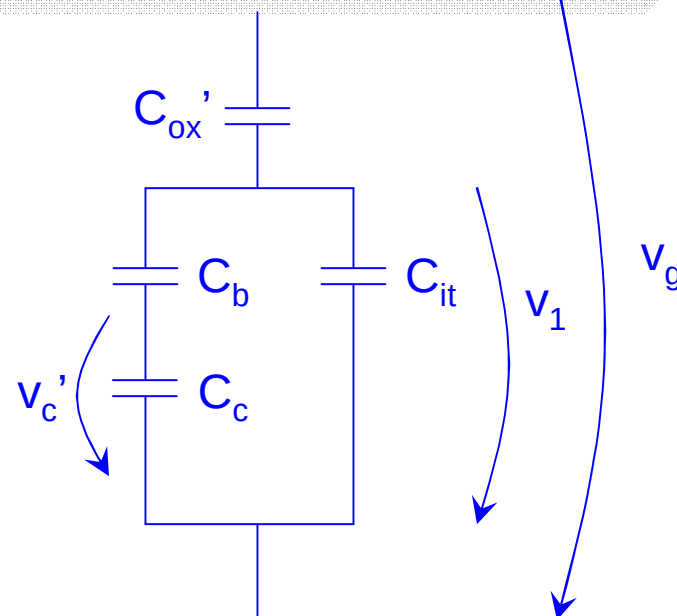
Heterostructure Barrier Layer: Case Studies

No barrier layer



C_{ox} : oxide capacitance
 C_c : channel capacitance
 C_{it} : D_{it} capacitance
 V_g : gate voltage
 V_c : channel voltage

With Barrier Layer



C_{ox}' : oxide capacitance
 C_c : channel capacitance
 C_b : barrier layer capacitance
 C_{it} : D_{it} capacitance
 V_g : gate voltage
 V_c' : channel voltage

A: Barrier Layer Case Study with $C_{ox}' = C_{ox}$

No barrier layer

$$\frac{V_c}{V_g} = \frac{C_{ox}}{C_c + C_{it} + C_{ox}}$$



Lower modulation efficiency v_c'/v_g for barrier layer case (D_{it} amplification and EOT increase)

With Barrier Layer

$$\frac{V_1}{V_g} = \frac{C_{ox}}{\frac{C_b C_c}{C_b + C_c} + C_{it} + C_{ox}}$$

$$\frac{V_c'}{V_1} = \frac{C_b}{C_b + C_c}$$

$$\frac{V_c'}{V_g} = \frac{C_{ox}}{C_c + \frac{C_b + C_c}{C_b} C_{it} + \frac{C_b + C_c}{C_b} C_{ox}}$$

Radosavljevic et al.: 1.8
Hill et al.: 2.8



D_{it} amplification (> 1) Increase in EOT

B: Barrier Layer Case Study with Identical EOT

$$C_{ox} \left[\begin{array}{c} C_{ox}' \\ C_b \end{array} \right] \quad \begin{array}{l} C_{ox}' \rightarrow \infty, C_b = C_{ox} \text{ (Schottky gate)} \\ C_b \gg C_{ox} \text{ (no barrier layer)} \end{array} \quad C_{ox}' = \frac{C_b C_{ox}}{C_b - C_{ox}}$$

No barrier layer

$$\frac{V_c}{V_g} = \frac{C_{ox}}{C_c + C_{it} + C_{ox}}$$

Higher modulation efficiency v_c'/v_g possible for barrier layer case but leakage current penalty

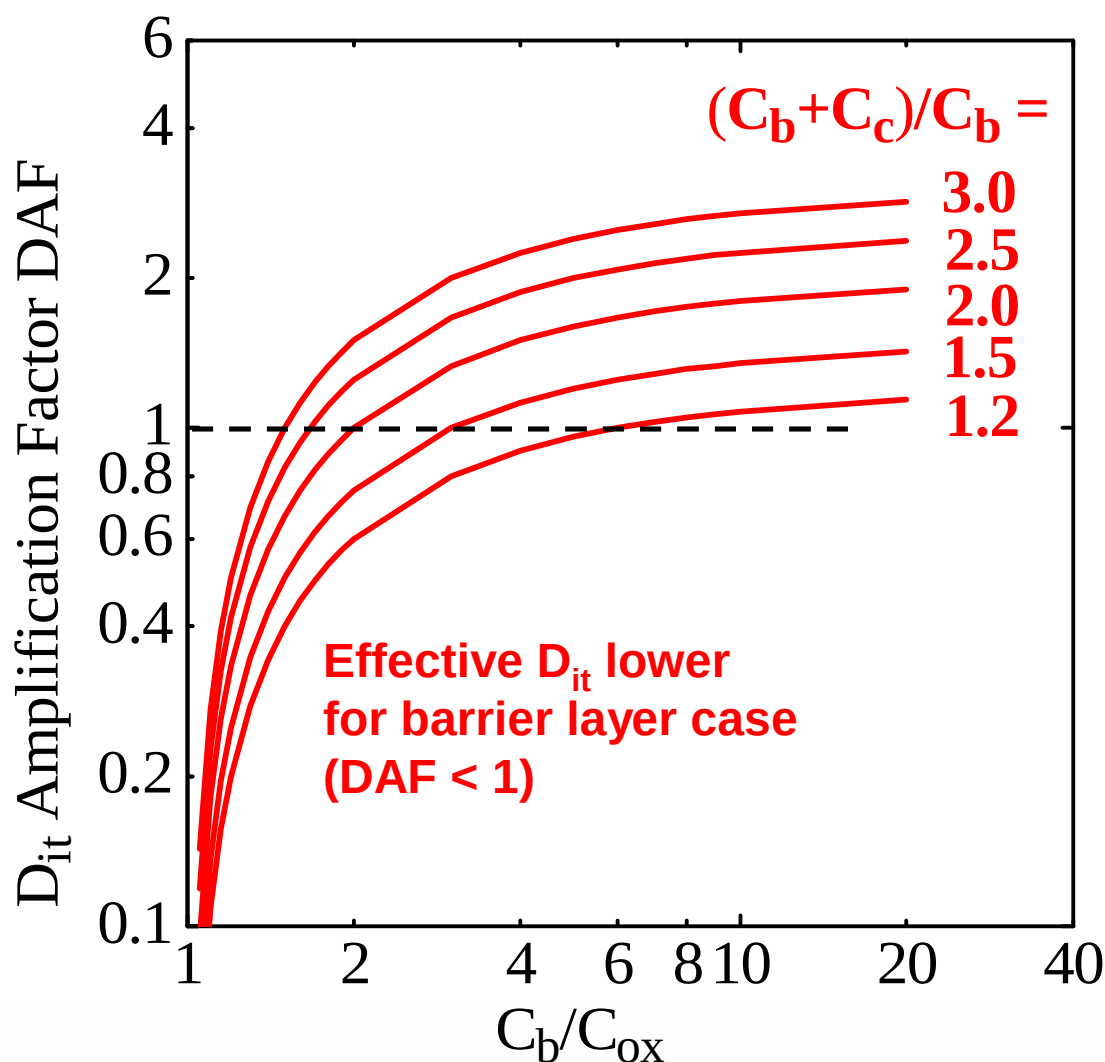
With Barrier Layer

$$\frac{V_c'}{V_g} = \frac{C_{ox}}{C_c + \left[\frac{C_b + C_c}{C_b} \right] \left[\frac{C_b - C_{ox}}{C_b} \right] C_{it} + C_{ox}}$$

$\uparrow$ > 1 $\uparrow$ ≤ 1
0 for $C_b = C_{ox}$
(Schottky gate limit)

D_{it} Amplification Factor DAF

Barrier Layer Case Study: Identical EOT



$$C_{ox} \left[\begin{array}{c} C_{ox}' \\ C_b \end{array} \right]$$

$C_{ox}' \rightarrow \infty, C_b = C_{ox}$
(Schottky gate)

C_{ox} : oxide capacitance without barrier layer
 C_{ox}' : oxide capacitance with barrier layer
 C_b : barrier layer capacitance

Further Reading

M. Passlack, J. Abrokwhah, and Z. Yu, "Self-aligned metal-oxide-compound semiconductor devices and methods of fabrication," US Patent 5,945,718, August 31, 1999.

M. Passlack, J. Wang, J. Abrokwhah, and Z. Yu, "Insulator-compound semiconductor interface structure," US Patent 6,359,294, March 19, 2002.

M. Passlack, K. Rajagopalan, J. Abrokwhah, and R. Droopad, "Flatband Mode MOSFET: "Implant-Free, High Mobility Flatband MOSFET: Principles of Operation," IEEE Trans. Electron. Dev., vol. 53, no. 10, pp. 2454-2459, 2006.

M. Passlack, R. Droopad, and G. Brammertz, "Suitability Study of Oxide/Gallium Arsenide Interfaces for MOSFET Applications," IEEE Trans. Electron. Dev., vol. ED-57, no. 11, pp. 2944-2956, 2010.